# A CANCELLATION NETWORK FOR FULL-DUPLEX FRONT END CIRCUIT

# Y. K. Chan and V. C. Koo

Faculty of Engineering & Technology Multimedia University Jalan Ayer Keroh Lama, Bukit Beruang, Melaka 75450, Malaysia

## B. K. Chung and H. T. Chuah

University Tungku Abdul Rahman 13, Jalan 13/6, Petaling Jaya, Selangor 46200, Malaysia

Abstract—A circulator is needed in a C-band airborne synthetic aperture radar system which employs single antenna configuration. The circulator provides full-duplex capability to transmit highpower RF signal and receive the echo signal via the same antenna simultaneously. Commercially available circulators with moderate isolation are inadequate for this application. An innovative Cancellation Network (CN) has been designed to enhance the performance of the conventional circulator. This paper highlights the conceptual design and measurement results of the CN. An improvement of more than 27 dB has been achieved.

# 1. INTRODUCTION

A circulator is a three-port non-reciprocal device that passes microwave energy in a forward direction but provides isolation in reverse direction. Circulator has been widely used in transmit and receive (T/R) modules of communication and radar system as a duplexer [1, 2]. Other applications include time delay switching application [3] and phase shifter [4]. Conventional ferrite circulators are constructed with permanent magnets and ferrite materials on microstrip circuit. The non-reciprocal action is brought by gyromagnetic action [1]. A new approach for realization of microwave circulator makes use of the nonreciprocal properties of microwave field effect transistor [5–7]. This

Corresponding author: Y. K. Chan (ykchan@mmu.edu.my).

active circulator can be implemented by Indium Gallium Arsenide (InGaAs)/Indium Aluminum Arsenide (InAlAs)/Indium Phosphide (InP) based microwave monolithic integrated circuit (MMIC) which offers small size and weight, and lower cost [7–11]. Recent development includes integrated active circulator antenna which combines hybrid active circulator with a passive microstrip antenna [12–14]. They are lightweight and low cost solutions for high-volume milimeter-wave system.

A number of articles on theory and design of three-port circulator can be found in the literature. The emphasis is on performance improvement in terms of insertion loss [15–17], wide band operation [20, 22, 28, 29], temperature stabilization [29, 30], isolation bandwidth [15, 16, 23], miniaturization [28, 30–32], isolation [17, 24] and high power operation [19, 21, 26, 30]. Isolation performance is critical in some radar systems. Insufficient isolation may result in the transmitter power leaking into the receiver input and causes saturation and intermodulation distortion to the receiver front-end circuit. In this paper a novel technique to improve isolation of a circulator is described. The circuit has been used on a C-band airborne synthetic aperture radar (SAR).

## 2. DESIGN THEORY AND IMPLEMENTATION

The basic block diagram of a radar system is shown in Fig. 1. Its major function is to isolate different part of an electronic system from one another. Poor isolation of the circulator may give rise to significantly large leakage signal from the transmitter to the sensitive receiver and interfere the reception of return echo.

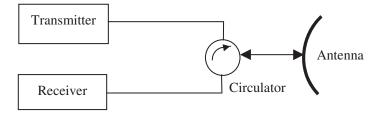


Figure 1. Radar system block diagram.

The block diagram of a circulator is shown in Fig. 2. With an input signal at port P1, most of the energy will appear at port P2, with small energy lost due to the insertion loss from P1 to P2. The coupling from P1 to P3 is undesirable and is characterized by the isolation of the circulator. On the other hand, a signal connected to

140

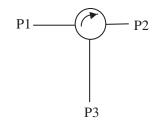


Figure 2. Component level diagram of a single circulator.

P2 will be routed to P3. For radar application, the transmitter is connected to P1, antenna to P2 and receiver to P3. This configuration allows simultaneous transmission and reception.

Commercial on-the-shelf circulators have moderate isolation over its operation bandwidth. It is possible to cancel the leakage signal by injecting a sample of the transmitter signal in equal amplitude but opposite phase into the receiver input. Fig. 3 shows the component level diagram of the cancellation network (CN) together with a circulator. The CN consists of 2 directional couplers, C1 and C2, a variable attenuator, R1, and two fixed-length coaxial cables. The sample of transmitter signal is coupled from main line by C1 and injected into the receiver by C2. The lengths of the cables are trimmed to provide the desired 180° phase shift and the variable attenuator controls the amplitude.

The CN has been implemented in an experimental airborne SAR sensor designed and developed at Multimedia University [33, 34], Malaysia. The airborne system is an inexpensive C-band, single polarization, linear-FM airborne radar sensor. The block diagram of the airborne SAR sensor is shown in Fig. 4.

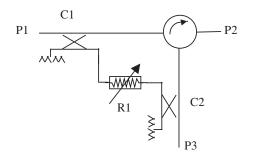


Figure 3. Component level diagram of a cancellation network.

Chan et al.

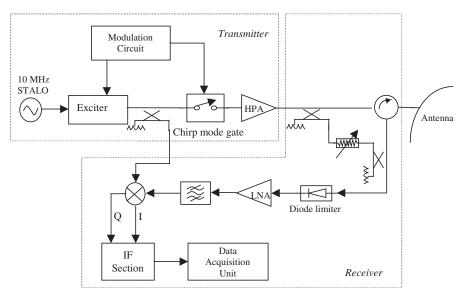


Figure 4. Component level diagram of the C-band SAR transmitter and receiver.

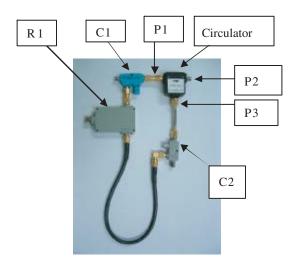


Figure 5. CN based on C-band circulator.

## 3. EXPERIMENTAL SETUP

In order to reduce the leakage signal from port P1 to port P3, the amplitude and phase of the signal coupled from CN to the receiver must

be properly adjusted. A commercial on-the-shelf C-band circulator was used. Fig. 5 shows the C-band circulator and the CN. The specification of the commercial circulator used in this experiment setup is list in the table below.

143

Model	H175FFF-S (Microwave Technology Corporation)
Frequency range	$5.5{-}6.5\mathrm{GHz}$
Isolation	$35\mathrm{dB}$
Insertion Loss	$0.3\mathrm{dB}$
VSWR	1.15:1 (max)
Power	10 W average, 100 W peak

 Table 1. Technical specification of circulator.

#### 3.1. Measurement of Leakage Signal from P1 to P3

The leakage signal can be measured by terminating both directional couplers with matched loads. The measurement setup is shown in Fig. 6. Both the phase and amplitude of  $S_{21}$  is measured over the operating bandwidth of our SAR system, i.e., 6 GHz center frequency with 20 MHz bandwidth. An Agilent Vector Network Analyzer (VNA) is used to perform the measurement.

Figure 7 shows the  $S_{21}$  of single circulator obtained from the measurement. It shows an average 35 dB isolation. Such isolation is not sufficient for the radar system which uses high power transmitter.

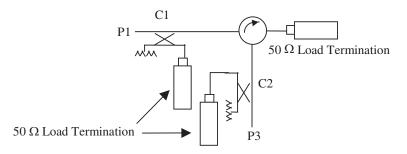


Figure 6.  $S_{21}$  measurement setup of leakage signal.

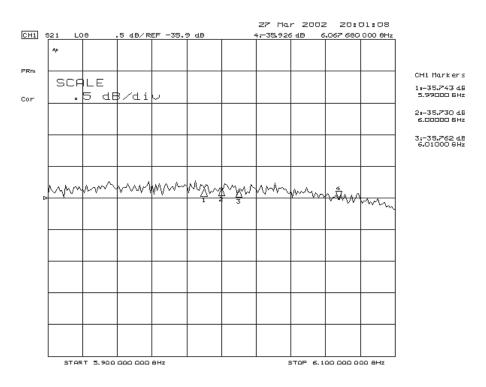


Figure 7.  $S_{21}$  of single circulator measured by VNA.

#### 3.2. Measurement of Injected Signal

In order to cancel the leakage signal, the RF cables must be properly selected so that the electrical length of the cables will produce a total phase shift of  $180^{\circ}$  compared to the leakage signal. The isolation of CN is measured using the setup shown in Fig. 8. The variable attenuator is adjusted until the  $S_{21}$  of the CN has the same magnitude as that of the single circulator. The length of the RF cables is trimmed to obtain the desired  $180^{\circ}$  phase difference.

The directional couplers are then connected to the circulator. The performance of CN is shown in Fig. 9, which indicates an excellent isolation of more than 62 dB throughout the 20 MHz bandwidth. This result represents an improvement of more than 27 dB. Fine-tuning of the variable attenuator is possible to achieve a better result.

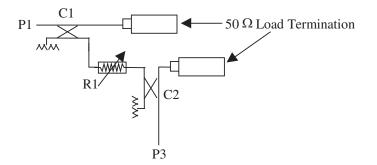


Figure 8. Measurement setup of injected signal by CN.

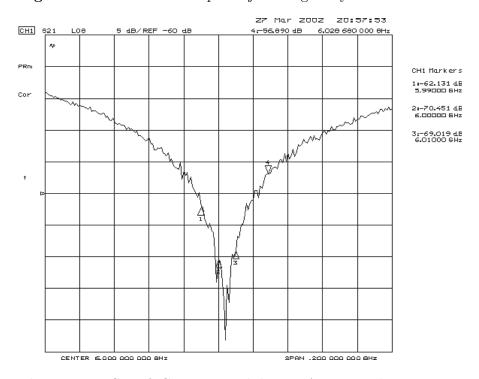


Figure 9.  $S_{21}$  of CN measured by VNA across the operation bandwidth of SAR sensor.

# 4. CONCLUSION

A new approach for improving isolation of circulator has been successfully demonstrated. This CN is incorporated in a prototype airborne SAR sensor developed by Multimedia University. It has been

145

shown that with proper injection of amplitude and phase, the leakage signal can be reduced significantly. An improvement of more than 27 dB can be achieved. The CN technique can be applied to other fullduplex communication systems to improve the isolation performance.

## ACKNOWLEDGMENT

This project is partially supported by the Malaysian Centre for Remote Sensing (MACRES), Malaysia.

## REFERENCES

- 1. Laverghetta, T. S., *Practical Microware*, Prentice-Hall, New Jersey, 1996.
- Pozar, D. M., *Microwave Engineering*, John Wiley & Sons, Inc., New York, 1998.
- Betts, F., D. H. Temme, and J. A. Weiss, "A switching circulator: S-band; Stripline; Remanent; 10 KW; 10 Microsecond; Temperature-stable," *G-MTT International Symposium Digest*, Vol. 66, No. 1, 275–280, May 1966.
- Silber, L. M. and A. Weis, "A fast switching high power C-band ferrite circulator," *IRE International Convention Record*, Vol. 12, 32–36, Mar. 1964.
- 5. Ayasli, Y., "Field effect transistor circulators," *Digests of Magnetics Conference*, AC2–AC2, Mar. 1989.
- Ayasli, Y., "Field effect transistor circulators," *IEEE Trans.* Magnetics, Vol. 25, No. 5, 3242–3247, Sep. 1989.
- 7. Ayasli, Y., "Non-ferrite non-reciprocal phase shifter and circulator," U. S. Patent, No. 4801901.
- Smith, M. A., "GaAs monolithic implementation of active circulators," *Microwave Symposium Digest*, Vol. 2, 1015–1016, May 1988.
- Hara, S., T. Tokumitsu, and M. Aikawa, "Novel unilateral circuits for MMIC circulators," *IEEE Trans. Microwave Theory Tech.*, Vol. 38, No. 10, 1399–1406, Oct. 1990.
- Kother, D., B. Hopf, T. Sporkmann, I. Wolff, and S. Kosslowski, "New types of MMIC circulators," *Microwave and Millimeter-Wave Monolithic Circuits Symposium*, 229–232, May 1995.
- 11. Berg, M., et al., "Active circulator MMIC in CPW technology using quarter micron InAlAs/InGaAs/InP in HFETs," *Eighth*

 $\mathbf{146}$ 

International Conference on Indium Phosphide and Related Materials, 68–71, Apr. 1996.

- Cryan, M. J. and P. S. Hall, "An integrated active circulator antenna," *IEEE Microwave and Guided Wave Letters*, Vol. 7, No. 7, 190–191, 1997.
- Gupta, S., P. K. Nath, and B. K. Sarkar, "Integrated microstrip shorted antenna using hybrid active circulator," Asia-Pacific Microwave Conference, 719–722, Dec. 2000.
- Kalialakis, C., M. J. Cryan, P. S. Hall, and P. Gardner, "Analysis and design of integrated active circulator antennas," *IEEE Trans. Microwave Theory Tech.*, Vol. 48, No. 6, 1017–1023, Jun. 2000.
- Miura, T., M. Kobayashi, and Y. Konishi, "Optimization of a lumped element circulator based on eigenvalue evaluation and structural improvement," *IEEE Trans. Microwave Theory Tech.*, Vol. 44, No. 12, 2648–2654, Dec. 1996.
- Fathy, A., E. Denlinger, D. Kalokitis, V. Pendrick, H. Johnson, A. Pique, K. S. Harshavardhan, and E. Belohoubek, "Miniature circulators for microwave superconducting systems," *IEEE MTT-S International Microwave Symposium Digest*, Vol. 1, 195–199, May 1995.
- Denlinger, E., R. Paglione, D. Kalokitis, E. Belohoubek, A. Pique, X. D. Wu, T. Venkatesan, A. Fathy, V. Pendrick, S. Green, and S. Mathews, "Superconducting nonreciprocal devices for microwave systems," *IEEE Microwave and Guided Wave Letters*, Vol. 2, No. 11, 449–451, Nov. 1992.
- Kadlec, J., "Tunable low-loss UHF circulator for cryogenic applications," *IEEE Trans. Microwave Theory Tech.*, Vol. 82, No. 2, 173–176, Feb. 1982.
- Helszajn, J. and M. E. Powlesland, "Low-loss high-peak-power microstrip circulators," *IEEE Trans. Microwave Theory Tech.*, Vol. 29, No. 6, 572–578, Jun. 1981.
- Miyoshi, T. and S. Miyauchi, "The design of planar circulators for wide-band operation," *IEEE Trans. Microwave Theory Tech.*, Vol. 28, No. 3, 210–214, Mar. 1980.
- Okada, F. and K. Ohwi, "Design of a high-power CW Y-junction waveguide circulator," *IEEE Trans. Microwave Theory Tech.*, Vol. 26, No. 5, 364–369, May 1978.
- Koning, J. G., Jr. R. J. Hamilton, and T. L. Hierl, "Full band low loss continuous tracking circulation in K-band," *IEEE Trans. Microwave Theory Tech.*, Vol. 25, No. 2, 152–155, Feb. 1977.
- 23. Okean, H. C. and L. J. Steffek, "Low loss, 3MM junction

circulator," *G-MTT International Microwave Symposium Digest*, Vol. 73, No. 1, 80–82, Jun. 1973.

- 24. Miura, T., "An experimental high isolation ferrite substrate circulator magnetized by trigonally symmetric pole pieces," *IEEE Trans. Magnetics*, Vol. 8, No. 3, 509–510, Sep. 1972.
- Edrich, J. and R. West, "Very low loss L-band circulator with gallium substituted YIG," *IEEE Trans. Magnetics*, Vol. 5, No. 3, 481–482, Sep. 1969.
- Buehler, G. V. and A. F. Eikenberg, "A VHF high-power Ycirculator," *IEEE Trans. Microwave Theory Tech.*, Vol. 9, No. 6, 569–570, Nov. 1961.
- 27. Arams, F., B. Peyton, and B. Kaplan, "Octave-bandwidth UHF/L-band circulator," *IEEE Trans. Microwave Theory Tech.*, Vol. 9, No. 3, 212–216, May 1961.
- Schloemann, E., "Miniature circulators," *IEEE Trans. Magnetics*, Vol. 25, No. 5, 3236–3241, Sep. 1989.
- Katoh, H., "Temperature-stabilized 1.7-GHz broad-band lumpedelement circulator," *IEEE Trans. Microwave Theory Tech.*, Vol. 23, No. 8, 689–696, Aug. 1975.
- Brown, J. and J. Clark, "Miniaturized, temperature stable, coaxial Y-junction circulators," *IEEE Trans. Microwave Theory Tech.*, Vol. 9, No. 3, 267–269, May 1961.
- Poirer, A. L., "An integrated microstrip circulator," *IEEE Trans.* Microwave Theory Tech., Vol. 19, No. 7, 661–662, Jul. 1971.
- Dunn, V. E. and R. W. Roberts, "New design techniques for miniature VHF circulators," *G-MTT Symposium Digest*, Vol. 65, No. 1, 147–152, May 1965.
- Chan, Y. K., B. K. Chung, and H. T. Chuah, "Transmitter and receiver design of an experimental airborne synthetic aperture radar sensor," *Progress In Electromagnetic Research*, PIER 49, 203–218, 2004.
- 34. Chan, Y. K., "Transmitter and receiver design of an airborne synthetic aperture radar," Master Dissertation, Multimedia University, Malaysia, 2002.