

STUDY AND SIMULATION OF AN EDGE COUPLE SPLIT RING RESONATOR (EC-SRR) ON TRUNCATED PYRAMIDAL MICROWAVE ABSORBER

H. Nornikman^{1,*}, B. H. Ahmad¹, M. Z. A. Abdul Aziz¹, F. Malek², H. Imran¹, and A. R. Othman¹

¹Department of Telecommunication Engineering, Faculty of Electronic and Computer Engineering, Universiti Teknikal Malaysia Melaka, Malaysia

²School of Electrical System Engineering, Universiti Malaysia Perlis, No. 12 & 14, Jalan Satu, Taman Seberang Jaya Fasa 3, Kuala Perlis 02000, Perlis, Malaysia

Abstract—Split ring resonator (SRR) structure can potentially be incorporated onto the truncated pyramidal microwave absorber. This study considers three different patterns of edge couple split ring resonator (EC-SRR) designs. Each EC-SRR design is then placed onto the truncated pyramidal microwave absorber. Outer split gap dimension widths of the EC-SRR are varied, and the various S_{21} performances are compared. This EC-SRR truncated pyramidal microwave absorber is simulated using CST Microwave Studio simulation software. The study and simulation are performed in low frequency range (0.01 GHz to 1 GHz) as well as in microwave frequencies range (1 GHz to 20 GHz). Simulation results of this EC-SRR show improvement of reflection loss and S_{11} performance in the high frequency range of the pyramidal truncated microwave absorber.

1. INTRODUCTION

Microwave absorbers can eliminate the reflected signal and are used in radio frequency (RF) anechoic chamber. Carbon is the main element used in a microwave absorber, as it provides the best reflection loss performance due to its ability to absorb the microwave signal. ϵ (epsilon) is the absolute permittivity of the dielectric, which is a measure of the electrostatic energy stored within it and therefore

Received 6 March 2012, Accepted 6 April 2012, Scheduled 20 April 2012

* Corresponding author: Hassan Nornikman (nornikman84@yahoo.com).

dependent on the material. Dielectric constant is equivalent to relative permittivity (ϵ_r) or the absolute permittivity (ϵ) relative to the permittivity of free space (ϵ_0). Equation (1) shows the formula of permittivity of free space while Equation (2) shows the formula of the wavelength.

$$\epsilon = \epsilon_r \epsilon_0 \quad (1)$$

$$\lambda = \frac{c}{f\sqrt{\epsilon}} \quad (2)$$

where c = speed of light = $3 \times 10^8 \text{ms}^{-1}$.

There are many ways to increase the pyramidal microwave absorber performance such as using the material with high carbon, using the new hybrid pyramidal shape, making an array pyramidal microwave absorber and adding the metamaterial structure on the pyramid. In this work, split ring resonator (one of metamaterial structure types) is used to increase the microwave absorber performance. Other metamaterial structures that can be used are artificial magnetic conductor (AMC) [1], electromagnetic band gap (EBG) [2], and photonic band gap (PBG) [3].

In microwave application, there are many applications that use split ring resonator (SRR) design such as the design of metamaterial antennas, microwave absorbers, mixers, filters and oscillators [4]. Many papers show the improved performance by employing the SRR in the design [5–19]. SRR can be incorporated onto a microwave absorber [20]. It has the potential to increase the reflection loss or S_{11} results of the microwave absorber.

Metamaterial or left handed material is an artificial material that does not exist in the real nature [21]. Metamaterial has categorized structure or design that has simultaneously negative permeability and permittivity. The metamaterial structure can significantly miniaturize the design size. It also achieves better reflection loss or return loss performances than a normal design without metamaterial structure [22] and improves antenna gain [23, 24]. The metamaterial artificial left-handed materials (LHMs) were initially discovered by Veselago [25]. Only after 3 decades, researchers in the field manage to apply the materials in practical applications. The first artificial metamaterial was fabricated by Smith et al. in 2000 [26]. Their design is based on Pendry split ring resonator-base artificial negative magnetic permeability media [27]. Split ring resonators (SRRs) design is used to produce the negative magnetic permeability.

Commercial microwave absorbers are designed to be made of plastic foamed-based materials like polystyrene or polyurethane with impregnating carbon as its mixed material. Beside that, there are also many researches on alternative materials, such as carbon nanotube

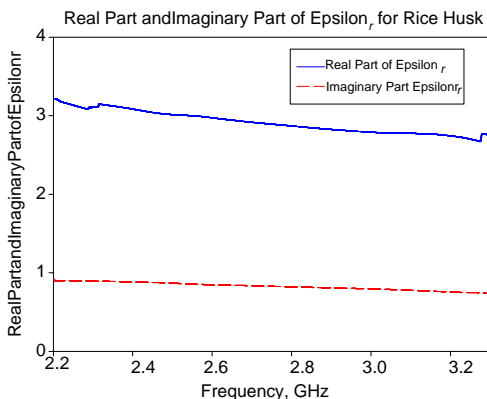


Figure 1. Real and imaginary parts of Epsilon_r (dielectric constant) for rice husk.

composite [28], ferrite absorber [29], ferroxide film [30] and other polymers materials. Agricultural residue like rice husk, a byproduct of paddy (*Oryza sativa*), has potential to apply as the based material in the microwave absorber fabrication. Figure 1 shows the dielectric constant that has been measured in the laboratory using dielectric probe technique. The dielectric constant for rice husk is $\epsilon_r = 2.9$. This dielectric constant has been measured using dielectric probe technique [31–33].

2. SPLIT RING RESONATOR (SRR) UNIT CELL

The split ring resonator can be applied in many designs, such as antennas [34, 35], microwave absorber [36–38], filters [39], frequency selective surface [40]. There are many types of split ring resonator, which have been designed by researchers, such as edge couple SRR (EC-SRR), broadside-couple SRR (BC-SRR), nonbianisotropic SRR (NB-SRR), Spiral SRR (S-SRR), open split ring resonator (O-SRR). Edge-coupled SRR (EC-SRR) was initially designed by Pendry in [41] and Smith et al. [42]. Since 1999, researchers have proposed many structures for metamaterial designs using split ring resonators (SRRs). These structures are used to produce the negative dielectric constant (permittivity) and negative permeability design. This structure can improve reflection loss (S_{11}) performances compared to the normal design [43].

Figure 2 shows the front view and ground view of the EC-SRR design. The front part of the EC-SRR has two rings, an inner ring

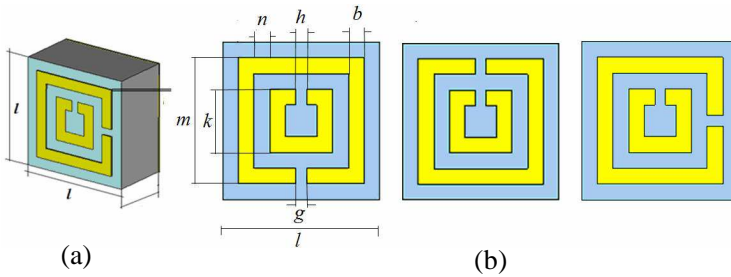


Figure 2. (a) The perspective view of single unit of EC-SRR structure. (b) Plan view of single unit of EC-SRR structure with three different pattern: Pattern *A*, Pattern *B*, and Pattern *C*. The SRR dimensions are as follows: Copper thickness, $t = 0.0035$ cm, Taconic *TLX 906207* substrate thickness, $t_{board} = 0.1615$ cm, side length, $l = 0.32$ cm, width of inner ring split gap, $h = 0.04$ cm, width of outer ring split gap, $g = 0.04$ cm, width of copper line, $b = 0.03$ cm, length of outer copper line, $m = 0.28$ cm, length of inner copper line, $k = 0.16$ cm and gap between inner and outer ring, $n = 0.02$ cm.

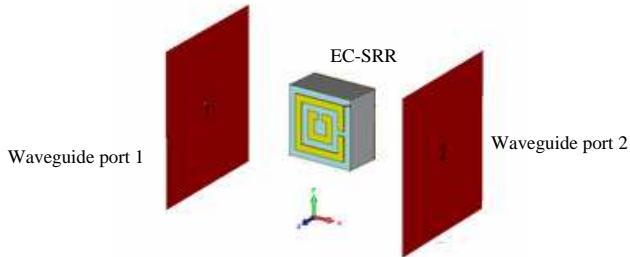


Figure 3. The simulation setup of single array of edge couple split ring resonator in CST Microwave Studio simulation software.

and outer ring, while the back part has a wire with dimension of $0.01 \text{ cm} \times 0.3 \text{ cm}$. Figure 3 shows the simulation setup run in CST Microwave Studio. The simulation consists of two sources of signal ports, namely waveguide port 1 and waveguide port 2. The EC-SRR is positioned right at the center between the two ports. CST Microwave Studio software is used to do the simulation to determine the performance of the EC-SRR on the microwave absorber. This EC-SRR is designed on Taconic *TLX 906207* board with dielectric constant of $\epsilon_r = 2.5$. The three patterns are differed by changing the length of gap between the outer rings of the SRR. The performance of

each SRR pattern, which has a different outer ring split gap width, is analyzed and compared. The gap values, g , chosen for comparison are 0.02 cm, 0.04 cm, 0.06 cm, and 0.08 cm, respectively. All patterns have the same dimension of 0.32 cm width \times 0.32 cm length and \times 0.1685 cm thickness.

There are many parameters that affect the performance or the result of the Split Ring Resonator such as the dimension of EC-SRR, N-number of SRR, split ring gap, thickness of the substrate used and others. In this study, three types of square shape EC-SRR patterns are designed to be incorporated into the truncated pyramidal microwave absorber.

3. THE TRUNCATED PYRAMIDAL MICROWAVE ABSORBER DESIGN

The design of the pyramidal microwave absorbers has been performed by Nornikman et al. [44–46]. Performances of three different stages of truncated pyramidal microwave absorber are studied. The stages are conventional truncated pyramidal microwave absorber stage, truncated pyramidal microwave absorber with single SRR stage, and truncated pyramidal microwave absorber with array SRR stage. Figure 4 shows the dimension of truncated pyramidal microwave absorber, which has two main elements, the pyramidal base part and the pyramidal body part. This paper discusses the designs of the normal truncated pyramidal microwave absorber, SRR truncated pyramidal absorbers, SRR array truncated pyramidal microwave absorber and reports the respective reflection losses. This microwave absorber is designed using

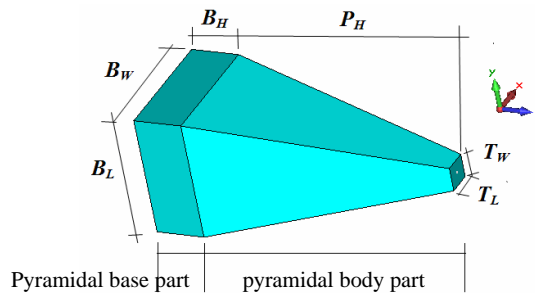


Figure 4. The dimensions of the truncated pyramidal microwave absorber; based length, $B_L = 5$ cm, based width, $B_W = 5$ cm, based height, $B_H = 2$ cm, pyramid height, $P_H = 13$ cm, top width, $T_W = 1$ cm, and top length, $T_L = 1$ cm.

CST Microwave Studio software.

Table 1 shows the differential dimension for the three different absorbers. TDK and VHP are commercial pyramidal microwave absorbers. They are currently available in the market. TDK absorber is designed by using a mixture of carbon and nonflammable material and VHP absorber by urethane foam. The truncated microwave absorber used in this study is made of agricultural waste, which is rice husk mixed with Urea Formaldehyde as its resin or bonding agent. It has a dielectric constant of $\epsilon_r = 2.9$.

Table 1. The differential dimension for TDK, VHP absorber and new design of truncated pyramidal microwave absorber.

Part	Symbol	Absorber dimension (cm)		
		TDK	VHP	Truncated
Pyramidal width	P_W	10	6.8	5
Pyramidal length	P_L	10	6.8	5
Pyramidal height	P_H	25	17.8	13
Base width	B_W	10	6.8	5
Based length	B_L	10	6.8	5
Based height	B_H	5	2.5	2
Top width	T_W	-	-	1
Top length	T_L	-	-	1

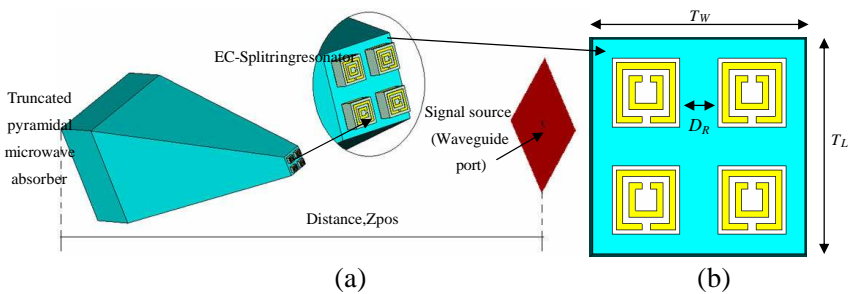


Figure 5. (a) The location of edge couple split ring resonator at the top of pyramidal microwave absorber (perspective view), Distance, $Z_{pos} = 30$ cm, (b) The array of EC-SRR structure (plan view), distance between two SRR, $D_r = 0.18$ cm, top length, $T_L = 1$ cm, top width $= T_W = 1$ cm.

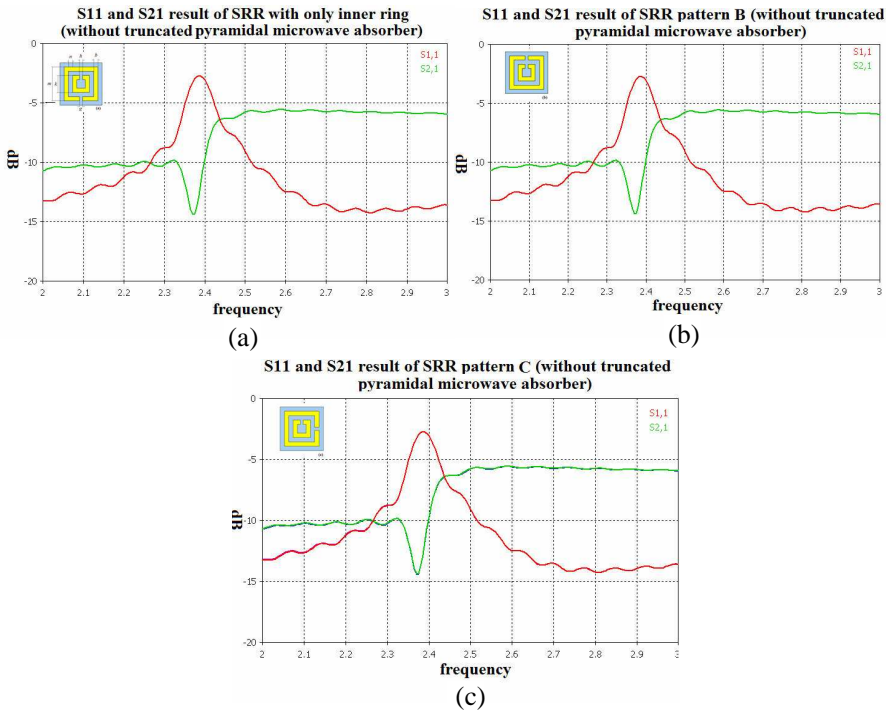


Figure 6. The S_{11} and S_{21} results for unit cell of EC-SRR designs (stand-alone SRR, without truncated pyramidal microwave absorbers): (a) SRR with only inner ring, (b) SRR Pattern A, (c) SRR Pattern B, (d) SRR Pattern C.

Basically, the dimension of this absorber is based on TDK ICT-030 and Eccosorb VHP-8-NRL but with modified dimension. Both TDK and VHP are pyramidal microwave absorbers with sharp corner at the top of the pyramid. In this work, the truncated pyramidal microwave absorber has 1 cm width \times 1 cm length surface at the top of the pyramid. The dimension of the truncated microwave absorber has been miniaturized as compared to the *TDK ICT-030* and *Eccosorb VHP-8-NRL* absorber. It significantly reduces the cost of mold fabrication.

Figure 5 shows the schematic diagram of the truncated pyramidal microwave absorber with the corresponding EC-SRR and source port location. The EC-SRR unit is placed at the top of the truncated pyramidal microwave absorber. The port is located at 30 cm away from the base absorber. The distance between pyramidal microwave absorber and source signal port, Z_{pos} , also affects the reflection loss performance of absorber [47].

4. RESULTS

4.1. S_{11} and S_{21} Result of Split Ring Resonator

Figure 6 shows S_{11} and S_{21} performance results of the SRR (stand-alone, without truncated pyramidal microwave absorber) for three different EC-SRR patterns (patterns *A*, *B* and *C*). The frequency ranges from 2 to 3 GHz. At 2.37 GHz, S_{21} result for pattern *A* is -14.418 dB, -14.421 dB for pattern *B*, and -14.381 dB for pattern *C*. It shows that all SRR patterns produce almost similar S_{21} and S_{11} results.

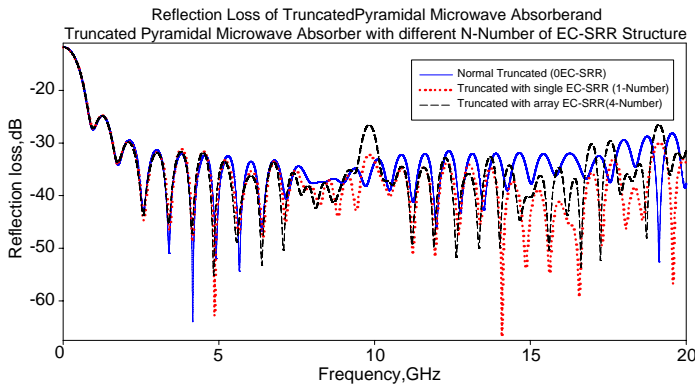


Figure 7. Reflection loss of normal truncated pyramidal microwave absorber and truncated pyramidal microwave absorber with EC-SRR structure split ring gap, $g = 0.04$ cm.

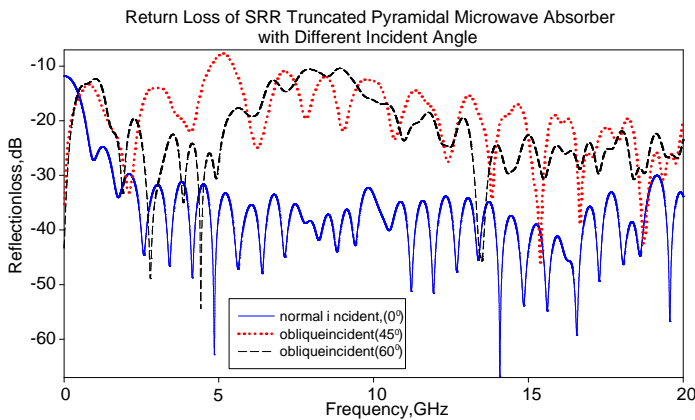


Figure 8. Reflection loss of normal truncated pyramidal microwave absorber with different incident angle. Split ring gap, $g = 0.04$ cm.

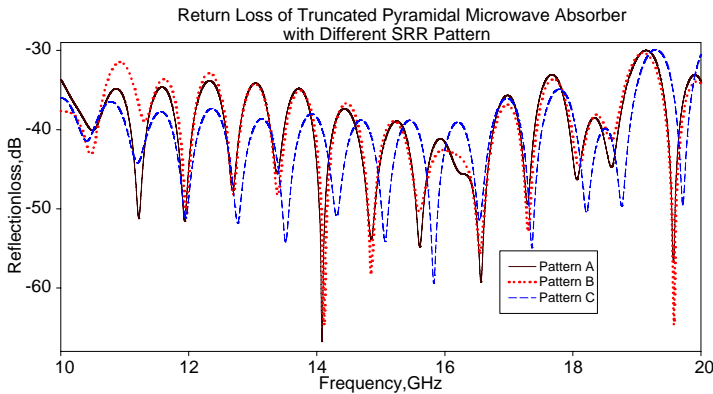


Figure 9. The reflection loss of truncated pyramidal microwave absorber with different SRR pattern. Split ring gap, $g = 0.04$ cm.

Table 2. Average reflection loss of conventional truncated pyramidal microwave absorber, absorber with single and EC-SRR array.

Frequency range (GHz)	Average reflection loss (dB)		
	Normal truncated absorber	Absorber with single EC-SRR	Absorber with EC-SRR array
0.01–1	-17.788	-17.724	-17.732
1–10	-35.870	-36.433	-36.170
10–15	-35.437	-39.582	-38.362
15–20	-33.469	-40.200	-36.017
0.01–20	-34.247	-37.219	-35.754

4.2. Effect of N-number of EC-SRR

Figure 7 and Table 2 show the reflection loss of conventional truncated pyramidal microwave absorber, absorber with one EC-SRR, and absorber with SRR array. Truncated pyramidal microwave absorber with single SRR has the best result among the three designs. The graph shows that the reflection loss of all three designs are better than -30 dB from the frequency range of 1.58 GHz to 20 GHz. The best point for this design is shown by absorber with single EC-SRR at frequency of 5.29 GHz with -66.774 dB. All three designs of the microwave absorber show similar performances in the low frequency range of 0.1 GHz to 1 GHz. In this frequency range it shows the worse reflection loss performance than the other frequency ranges.

Table 3. Average reflection loss of truncated pyramidal absorber with different incident angle.

Frequency range (GHz)	Average reflection loss (dB)		
	Normal incident (0°)	Oblique incident (45°)	Oblique incident (60°)
0.01–1	–17.724	–17.737	–18.392
1–10	–36.433	–16.541	–19.762
10–15	–39.582	–19.329	–23.871
15–20	–40.200	–25.967	–26.066
0.01–20	–37.219	–19.664	–22.311

Table 4. The average reflection loss of truncated pyramidal microwave absorber with different SRR pattern.

Frequency range (GHz)	Average reflection loss (dB)			
	No SRR structure	SRR pattern <i>A</i>	SRR pattern <i>B</i>	SRR pattern <i>C</i>
0.01–1	–17.788	–17.724	–17.735	–17.876
1–10	–35.870	–36.433	–36.664	–36.868
10–15	–35.437	–39.582	–38.900	–41.404
15–20	–33.469	–40.200	–40.305	–40.742
0.01–20	–34.247	–37.219	–37.176	–38.008

4.3. Effect of Incident Angle

Figure 8 and Table 3 show the return loss of SRR truncated pyramidal microwave absorber with different incident angles. Three different incident angles are considered in this section, i.e., normal incident of 0° or straight line, 45° of oblique angle, and 60° of oblique angle. It is shown that the normal incident achieves the better reflection loss performance than the other incident angles. The average return losses (between 0.01 GHz and 20 GHz) for 450 and 600 are only -19.664 dB and -22.311 dB compared to the normal incident with -37.219 dB.

4.4. Effect of EC-SRR Pattern

Figure 9 and Table 4 show the reflection loss for pyramidal microwave absorber with different SRR patterns. In the frequencies ranging from 1 GHz to 10 GHz, S_{11} value for the SRR truncated pyramidal microwave absorbers is around 1 to 2% larger than the result obtained

for the conventional truncated pyramidal microwave absorber. For the frequencies ranging from 10 GHz to 15 GHz, better overall reflection loss performances are observed than the conventional truncated pyramidal microwave absorber. SRR pattern *A* shows a 9.38% larger result, SRR pattern *B* a 7.65% larger result, and SRR pattern *C* a 13.98% larger result.

For the high frequencies ranging from 15 GHz to 20 GHz, larger reflection loss values for the three SRR patterns are observed than the results obtained using the conventional truncated pyramidal microwave absorber. For SRR pattern *A*, the reflection loss result is larger by 20.11%, for SRR pattern *B* 20.43% larger, and for SRR pattern *C* 21.73% larger. For comparison, SRR pattern *C* shows the best performance among all the analyzed SRR patterns. Overall, in the frequencies ranging from 0.01 to 20 GHz, EC-SRR pattern *C* achieves -38.008 dB of reflection loss result compared to the conventional truncated pyramidal microwave absorber which achieves -34.247 dB of reflection loss result.

4.5. Effect of Split Ring Gap

Figure 10 and Table 5 show the reflection loss of the truncated pyramidal microwave absorber of SRR with different split gap dimensions, g values. From Table 5, it can be seen that the minimum value of the split gap ($g = 0.02$ cm) provides a better reflection loss result performance than the other wider ring gap dimensions. It achieves an average return loss of -37.432 dB for 0.01 GHz to 20 GHz.

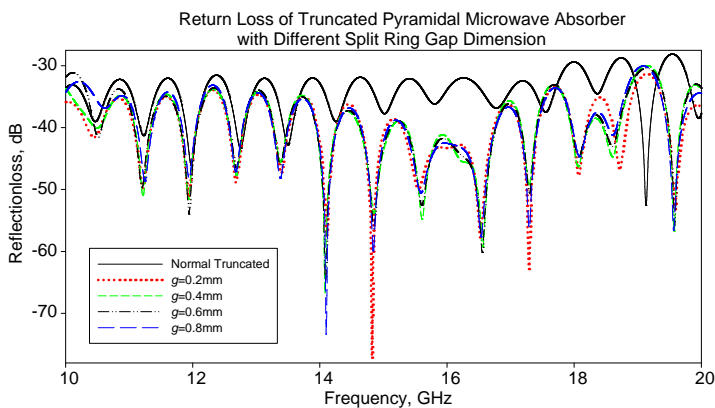


Figure 10. The reflection loss of truncated pyramidal microwave absorber with different ring gap dimension.

Table 5. The average reflection loss of truncated pyramidal microwave absorber with different ring gap dimension.

Frequency range (GHz)	Average reflection loss with ring gap dimension, g (cm)			
	0.02 cm	0.04 cm	0.06 cm	0.08 cm
0.01–1	–17.685	–17.724	–17.716	–17.732
1–10	–36.518	–36.433	–36.598	–36.727
10–15	–39.847	–39.582	–39.234	–39.033
15– 0	–40.637	–40.200	–40.146	–40.286
1–20	–37.432	–37.219	–37.195	–37.237

5. CONCLUSION

An edge-couple split ring resonator (EC-SRR) design incorporated into the truncated pyramidal microwave absorber has been presented in this paper. The incorporation of the EC-SRR increases the reflection loss performance of the truncated pyramidal microwave absorber, especially in high frequency range (15 GHz to 20 GHz). It is also shown that pattern C of the EC-SRR design has the best results amongst all the three EC-SRR designs analyzed. The outer split ring gap width of $g = 0.02$ (i.e., the minimum value of split gap analyzed) results in the best reflection loss performance.

REFERENCES

1. Al-Hasan, M. J., T. A. Denidni, and A. Sebak, “A new UC-EBG based-dielectric resonator antenna for millimeter-wave applications,” *2011 IEEE International Symposium on Antennas and Propagation (APSURSI)*, 1274–1276, 2011.
2. Elsheakh, D. N., H. A. Elsadek, E. A. Abdallah, M. F. Iskander, and H. Elhenawy, “Ultrawide bandwidth umbrella-shaped microstrip monopole antenna using spiral artificial magnetic conductor (SAMC),” *IEEE Antennas and Wireless Propagation Letters*, Vol. 8, 1255–1258, 2009.
3. Qiang, W. Y. and F. Tao, “The study on a patch antenna with PBG structure,” *Third International Symposium on Intelligent Information Technology Application (IITA 2009)*, Vol. 3, 565–567, 2009.

4. Chang, K., *Microwave Ring Circuit and Antennas*, John Wiley, New York, 1996.
5. Veselago, V. G., "The electrodynamics of substances with simultaneously negative values of ϵ and μ ," *Soviet Physics Uspekhi*, Vol. 10, No. 4, 509–514, 1968.
6. Katsarakis, N., T. Koschny, and M. Kafesaki, "Electric coupling to the magnetic resonance of split ring resonators," *Applied Physics Letters*, Vol. 84, No. 15, Apr. 12, 2004.
7. Wu, B., B. Li, T. Su, and C.-H. Liang, "Study on transmission characteristic of split ring resonator defect ground structure," *PIERS Online*, Vol. 2, No. 6, 710–714, 2006.
8. Garcia-Garcia, J., F. Aznar, M. Gil, J. Bonache, and F. Martin, "Size reduction of SRRs for metamaterial and left handed media design," *PIERS Online*, Vol 3, No. 3, 266–269, 2007.
9. Niu, J.-X., X.-L. Zhou, and L.-S. Wu, "Analysis and application of a novel structures based on split ring resonators and coupled lines," *Progress In Electromagnetics Research*, Vol. 75, 153–162, 2007.
10. Niu, J.-X. and X.-L. Zhou, "Analysis of balanced composite right/left handed structure based on different dimensions of complementary split ring resonators," *Progress In Electromagnetics Research*, Vol. 74, 341–351, 2007.
11. Nornikman, H., F. Malek, P. J. Soh, and A. A. H. Azremi, "Design a rice husk pyramidal microwave absorber with split ring resonator," *The Asia-Pacific Symposium on Applied Electromagnetics and Mechanics 2010 (APSAEM 2010)*, 2010.
12. Rahim, M. K. A., H. A. Majid, and T. Masri, "Microstrip antenna incorporated with left-handed metamaterial at 2.7 GHz," *IEEE International Workshop on Antenna Technology (iWAT 2009)*, 1–4, 2009.
13. Ezanuddin, A. A. M., F. Malek, and P. J. Soh, "Investigation of complementary split ring resonator with dielectric ring," *Loughborough Antennas and Propagation Conference (LAPC)*, 297–300, 2010.
14. Yuandan, D. and T. Itoh, "Miniaturized patch antennas loaded with complementary split-ring resonators and reactive impedance surface," *5th European Conference on Antennas and Propagation (EUCAP)*, 2415–2418, 2011.
15. Quevedo-Teruel, O., M. N. M. Kehn, and E. Rajo-Iglesias, "Dual-band patch antennas based on short-circuited split ring resonators," *IEEE Transactions on Antennas and Propagation*,

- Vol. 59, No. 8, 2758–2765, 2011.
16. Jiun-Peng, C. and H. Powen, “A miniaturized slot dipole antenna capacitively fed by a CPW With split ring resonators,” *2011 IEEE International Symposium on Antennas and Propagation (APSURSI)*, 779–781, 2011.
 17. Lin, H.-H., C.-Y. Wu, and S.-H. Yeh, “Metamaterial enhanced high gain antenna for WiMAX application,” *2007 IEEE Region 10 Conference (TENCON 2007)*, 1–3, 2007.
 18. Majid, H. A., M. Rahim, and T. Masri, “Left handed metamaterial design for microstrip antenna application,” *IEEE International RF and Microwave Conference (RFM 2008)*, 218–221, 2008.
 19. Lai, X., Q. Li, P.-Y. Qin, B. Wu, and C.-H. Liang, “A novel wideband bandpass filter based on complementary split-ring resonator,” *Progress In Electromagnetics Research C*, Vol. 1, 177–184, 2008.
 20. Bilotti, F. and L. Vegni, *Design of Metamaterial-Based Resonant Microwave Absorbers with Reduced Thickness and Absence of Metallic Backing*, Springer Sciences — Business Media B. V., 2009.
 21. Rahim, M. K. A., H. A. Majid, and T. Masri, “Microstrip antenna incorporated with left-handed metamaterial at 2.7 GHz,” *IEEE International Workshop on Antenna Technology (iWAT 2009)*, 1–4, 2009.
 22. Feresidis, A. and J. C. Vardaxoglou, “Flat plate millimetre wave antenna based on partially reflective FSS,” *International Conference on Antennas and Propagation*, Vol. 1, 33–36, 2001.
 23. Liu, S.-H., C.-H. Liang, W.-Ding, L.-Chen, and W.-T. Pan, “Electromagnetic wave propagation through a slant waveguide of uniaxially anisotropic dispersive metamaterial,” *Progress In Electromagnetics Research*, Vol. 76, 467–475, 2007.
 24. Hwang, R.-B., H.-W. Liu, and C. Y. Chin, “A metamaterial based E-plane horn antenna,” *Progress In Electromagnetics Research*, Vol. 93, 275–289, 2009.
 25. Veselago, V. G., “The Electrodynamics of substances with simultaneously negative values of permittivity and permeability,” *Soviet Physics USPEKI*, Vol. 10, No. 4, 509–514, 1968.
 26. Smith, D. R., W. J. Padilla, D. C. Vier, S. C. Nemat-Nasser, and S. Schultz, “Composite medium with simultaneously negative permeability and permittivity,” *Phys. Rev. Lett.*, Vol. 84, 4184–4187, 2000.

27. Pendry, J. B., A. J. Holden, D. J. Robbins, and W. J. Stewart, "Magnetism from conductors and enhanced nonlinear phenomena," *IEEE Trans. Microwave Theory Tech.*, Vol. 47, 2075–2084, 1999.
28. Saib, A., L. Bednarz, R. Daussin, C. Bailly, X. Lou, J.-M. Thomassin, C. Pagnouille, C. Detrembleur, and R. Jerome, "Carbon nanotube composites for broadband microwave absorbing materials," *2005 European Microwave Conference*, Vol. 1, 2005.
29. Kotsuka, Y. and H. Yamazaki, "Fundamental investigation on a weakly magnetized ferrite absorber," *IEEE Transaction on Electromagnetic Compatibility*, Vol. 42, No. 2, 116–124, 2000.
30. Nedkov, I., L. Milenova, and N. Dishovsky, "Microwave polymer-ferroxide film absorbers," *IEEE Transactions on Magnetics*, Vol. 30, No. 6, 4545–4547, 1994.
31. Nornikman, H., F. B. A. Malek, P. J. Soh, A. A. H. Azremi, F. H. Wee, and A. Hasnain, "Parametric study of pyramidal microwave absorber using rice husk," *Progress In Electromagnetics Research*, Vol. 104, 145–166, 2010.
32. Nornikman, H., F. Malek, M. Ahmed, F. H. Wee, P. J. Soh, A. A. H. Azremi, S. A. Ghani, A. Hasnain, and M. N. Taib, "Setup and results of pyramidal microwave absorbers using rice husks," *Progress In Electromagnetics Research*, Vol. 111, 141–161, 2011.
33. Malek, M., E. M. Cheng, O. Nadiah, H. Nornikman, M. Ahmed, M. Z. A. Abd Aziz, A. R. Othman, P. J. Soh, A. A. H. Azremi, A. Hasnain, and M. N. Taib, "Rubber tire dust-rice husk pyramidal microwave absorber," *Progress In Electromagnetics Research*, Vol. 117, 449–447, 2011.
34. Ezanuddin, A. A. M., F. Malek, and P. J. Soh, "Investigation of complementary split ring resonators with dielectric ring," *2010 Loughborough Antennas and Propagation Conference (LAPC)*, 297–300, 2010.
35. Majid, H. A., M. Rahim, and T. Masri, "Left handed metamaterial design for microstrip antenna application," *IEEE International RF and Microwave Conference (RFM 2008)*, 218–221, 2008.
36. Huang, L. and H. Chen, "Multi-band and polarization insensitive metamaterial absorber," *Progress In Electromagnetics Research*, Vol. 113, 103–110, 2011.
37. Zhu, B., Z. Wang, C. Huang, Y. Feng, J. Zhao, and T. Jiang, "Polarization insensitive metamaterial absorber with incident angle," *Progress In Electromagnetics Research*, Vol. 101, 231–239,

- 2010.
38. Wang, J., S. Qu, Z. Fu, H. Ma, Y. Yang, X. Wu, Z. Xu, and M. Hao, "Three-dimensional metamaterial microwave absorbers composed of coplanar magnetics and electric resonators," *Progress In Electromagnetics Research Letters*, Vol. 7, 15–24, 2009.
 39. Das, S., A. Kundu, S. Maity, S. Dhar, and B. Gupta, "Novel compact CPW filter for MICs using metamaterial structures", *2011 11th Mediterranean Microwave Symposium (MMS)*, 286–289, 2011.
 40. Kern, D. J., D. H. Werner, A. Monorchio, L. Lanuzza, and M. J. Wilhelm, "The Design synthesis of multiband artificial magnetic conductors using high impedance frequency selective surfaces," *IEEE Transactions on Antennas and Propagation*, Vol. 53, No. 1, Part 1, 8–17, 2005.
 41. Pendry, J. B., "Negative refraction makes a perfect lens," *Physical Review Letters*, Vol. 85, No. 18, 3966–3969, 2000.
 42. Smith, D. R., W. Padilla, D. C. Vier, S. C. Nemat-Nasser, and S. Schultz, "Negative permeability from split ring resonator arrays," *2000 Conference on Lasers and Electro-Optics Europe*, 2000.
 43. Nornikman, H., F. Malek, P. J. Soh, and A. A. H. Azremi, "Design a rice husk pyramidal microwave absorber with split ring resonator," *The Asia-Pacific Symposium on Applied Electromagnetics and Mechanics 2010 (APSAEM 2010)*, 2010.
 44. Nornikman, H., P. J. Soh, and A. A. H. Azremi, "Performance simulation of pyramidal and wedge microwave absorbers," *3rd Asian Modelling Symposium (AMS 2009)*, 649–654, 2009.
 45. Nornikman, H., P. J. Soh, and A. A. H. Azremi, "Modelling simulation stage of pyramidal and wedge absorber microwave absorber design," *4th International Conference on Electromagnetic Near Field Characterization and Imaging (ICONIC'09)*, 2009.
 46. Nornikman, H., P. J. Soh, A. A. H. Azremi, F. H. Wee, and F. M. Malek, "Investigation of an agricultural waste as an alternative material for microwave absorber," *PIERS Online*, Vol. 5, No. 6, 506–510, 2009.
 47. Nornikman, H., Malek, F., P. J. Soh, and A. A. H. Azremi, "Effect on source signal condition for pyramidal microwave absorber performance," *International Conference on Computer & Communication Engineering (ICCCE 2010)*, 289–293, 2010.