# Unilateral Mini NMR Sensor Used for Assessing the Aging Status of the Sheds of Composite Insulators

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Abstract—The number of composite insulators used in power transmission lines increases year by year. To detect and assess the aging status accurately concerns the security and stability of the power system. In order to achieve nondestructive testing of the sheds of composite insulators, a unilateral mini Nuclear Magnetic Resonance (NMR) sensor is proposed in this paper. The design of the magnet body and the optimization of the RF coil are presented. The Carr-Purcell-Meiboom-Gill (CPMG) sequence was employed to record the <sup>1</sup>H relaxation curves of the sheds of three composite insulators from 110 kv lines with different service years. The curves were fitted to both single exponential function and inverse Laplace transformation functions. The results demonstrate that an increase of service year of the insulator results in a decrease of the effective transverse relaxation time  $(T_{2eff})$ . It is indicated that the sensor has a potential to assess the aging status of the composite insulators.

# 1. INTRODUCTION

Due to the light weight, excellent anti-flashover performance and high mechanical strength, composite insulators [1, 2] have been widely used in high voltage transmission lines. Composite insulators are typically composed of a central rod made of fiber reinforced plastic and an outer weather shed made of silicone rubber. The sheds therein are used for protecting the central rod from the moisture in the air and providing a sufficiently long leakage current path. In service under the power grid overvoltage and the corrosion of the natural environment [3, 4] (pollution, acid rain, high temperature, etc.), the surfaces of the sheds become sclerous and chalked over time and the hydrophobicity on which degrades. These phenomena will cause the decline of insulating performance of composite insulator sheds and frequent creeping accidents thus threatening the stable operation of the power system. Assessing the aging status of the sheds accurately and replacing the over-aged composite insulators in time are therefore critical to ensuring the safe and stable operation of the power system.

Several methods have been reported for assessing the aging status [5] of the sheds of composite insulators, such as direct observation method [6,7], water spray classification [8–10], leakage current method [11–13], contact angle method [14], thermally stimulated current method [15–17], etc. Direct observation method and water spray classification are easily influenced by the subjective judgment of maintenance personnel and therefore unstable. Leakage current method is also unreliable since the factors affecting the current are complicated and random. Contact angle method and thermally stimulated current method are destructive and generally used for off-line measurements in laboratory. So far, there is no nondestructive method to assess the aging status of the composite insulator sheds accurately in practical engineering.

Traditional NMR technologies have been used to investigate the aging and degradation of polymer materials in laboratory [18]. Due to its portability and non-destructive, portable low-field NMR has been

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successfully employed for detecting the aging status of rubber materials [19] in practical engineering. In [20], our team has introduced a portable low-field NMR sensor for assessing the aging status of the sheds of composite insulators at project site. The measurement results show that the sensor can successfully distinguish the aging status of composite insulators with different service years. However, the static magnetic field  $B_0$  of the sensor is generated by two magnet discs (Figure 1), and the distance between the two magnets needs to be adjusted according to the thickness of the sheds of different insulators. The RF circuit therefore has to be retuned since the distribution of  $B_0$  changes with the distance, and it increases the difficulty of the measurement at project site. In addition, the  $B_0$  is very inhomogeneous which results in low signal to noise ratio (SNR).

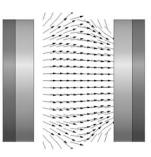


Figure 1. The structure of the magnet body of the portable low-field NMR sensor in [20]. The magnet body consists of two magnets, and the  $B_0$  is between them.

Since the aging status of the upper surface of the sheds is higher than the lower surface, the upper surfaces are more interesting for the measurements. To solve the above drawbacks of the portable lowfield NMR sensor, we propose a unilateral mini NMR (UMR) sensor in this paper for investigating the aging status of the composite insulators. This UMR sensor has more uniform  $B_0$  and does not need to adjust its structure in the measurements. To obtain high SNR, we also improve the structure of the RF coil. In the following text, the structure of the sensor, measurements on composite insulators and the results are described in detail.

## 2. UMR SENSOR

The sensor (Figure 2(a)) mainly consists of three cubic magnets, an aluminum shell and an RF coil with its tuning and matching circuit. The magnets are used to generate the static magnetic field  $B_0$ . The aluminum shell is to assemble the magnets, and the RF coil with its tuning and matching circuit is to transmit the excitation pulses and receive the NMR signals of the samples. An illustration of the UMR sensor testing the composite insulator sheds is shown in Figure 2(b).

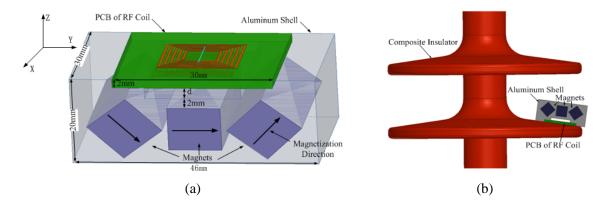
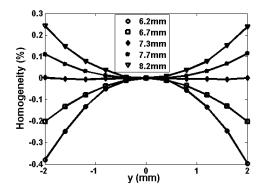


Figure 2. (a) The diagram of the UMR sensor, (b) an illustration of the UMR sensor testing the composite insulator sheds.

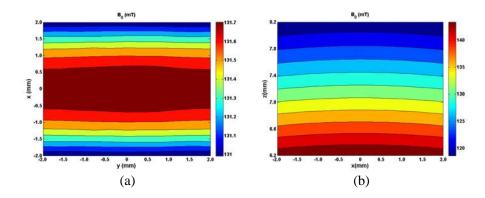
#### 2.1. Structure of the Magnet Body

It is well known that more uniform  $B_0$  means more hydrogen excited so that higher SNR can be obtained. The aim of the design of magnet structure is to generate a relatively homogeneous  $B_0$ . Meanwhile, for a non-destructive measurement at project site, we must take the portability into account during the design process of the sensor. In this work, the magnet structure (Figure 2(a)) of the sensor was designed to consist of three magnet blocks (SmCo YXG-32L, Shanghai CJ Magnet Industry Co., Ltd.) with the same size. In consideration of the distance between the two adjacent sheds and the size of sheds, the dimension of each magnet was set as  $10 \text{ mm}(y) \times 10 \text{ mm}(z) \times 30 \text{ mm}(x)$ .

As shown in Figure 2(a), the magnetization orientation of the magnets was fixed. The relative position of the three magnets was adjusted to optimize the homogeneity of the  $B_0$  in the region of interest (a 4 mm × 4 mm × 4 mm area which is 5 mm above the surface of the central magnet), and the optimal structure of the magnet body was determined after. The optimization process was described in the reference [21] in detail. For the optimal magnet structure, Figure 3 shows the variation of  $B_0$ along y axis at different z simulated by Maxwell 3D (Ansoft, Pittsburgh, PA, USA). It can be seen that the  $B_0$  along y axis at z = 7.3 mm is the most homogeneous. Thus, we chose the xy plane at z = 7.3 mm to be the optimal plane for the upper surface of the test sheds and the orientation of  $B_0$  on this plane is along y axis. Figure 4(a) shows the distribution of the simulated  $B_0$  on the optimal plane, and Figure 4(b) is on xz plane at y = 0 where the homogeneity is 0.5%.



**Figure 3.** The variation of the simulated  $B_0$  along y axis at different z when x = 0.



**Figure 4.** (a) The distribution of the simulated  $B_0$  on the optimal plane. (b) The distribution of the simulated  $B_0$  on xz plane at y = 0.

## 2.2. Design of RF Coil

As discussed above, the plane with most homogeneous  $B_0$  was 7.3 mm away from the surface of the central magnet. Due to the RF field  $B_1$  decreasing rapidly with the distance from the surface of the RF coil, the plane 0.5 mm above the coil was selected as the optimal plane for RF pulses, and the RF coil

was then placed 6.8 mm away from the central magnet. A flat rectangular spiral coil was employed to generate  $B_1$  perpendicular to  $B_0$  (along y axis). The  $B_0$  strength of the optimal plane was 117.83 mT, according to Larmor formula, and the excitation frequency was therefore set to be 5.017 MHz.

In NMR, when the other conditions such as the excited volume of the sample, the current through the RF coil and the static magnetic field  $B_0$  are the same, and the SNR can be expressed (Formula (1)) by the strength of  $B_1$  and the AC resistance of the coil [22]. Therefore, the performance of RF coil directly determines the quality of the NMR signals.

$$SNR \propto \frac{B_1}{\sqrt{R}}$$
 (1)

Usually in UMR, the SNR is much lower than in the traditional NMR. In order to get reliable and stable results, the SNR of the measurements is desired to be as high as possible. In 2.1, the magnet structure was arranged to get the most homogeneous  $B_0$  in the region of interest, and in this section the size and turns of the RF coil were adjusted to optimize the SNR. The distance between two adjacent wires and the width of the wires were fixed to be 0.5 mm.

Since there is no simple analytical relation among  $B_1$ , R and the structural parameters of the RF coil, we can only calculate  $B_1$  and R to optimize the structure of coil by simulating in Maxwell 3D separately. Figure 5 shows  $B_1$  distribution on the optimal plane of three RF coil with different parameters. According to Formula (1), the SNR is proportional to the ratio of  $B_1$  to square root of R. Therefore, the relative SNR of each coil can be calculated from the simulation results. Because  $B_1$  is symmetrically distributed on the optimal plane, Figure 6 only shows the relative SNR of the three coils along x axis. Comparing both the strength and homogeneity of  $B_1$ , it can be observed that  $B_1$  in Figure 5(b) is the best. The parameters of the optimal coil are as follows: the size is 12.5 mm × 12.5 mm; the coil has two layers with 5 turns at each layer; the thickness of the PCB is 2 mm.

Figure 7 shows the orientations of  $B_0$  and  $B_1$  on the optimal plane, and most of  $B_0$  are perpendicular to  $B_1$  which means  $B_1$  efficiently used.

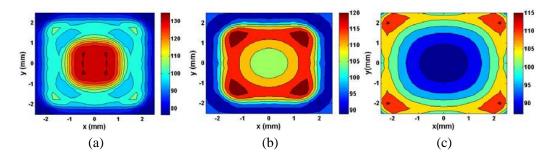


Figure 5. RF field distribution on the optimal plane of three RF coils of different size with the optimal turns, (a)  $11 \text{ mm} \times 11 \text{ mm}$ , (b)  $12.5 \text{ mm} \times 12.5 \text{ mm}$ , (c)  $14 \text{ mm} \times 14 \text{ mm}$ .

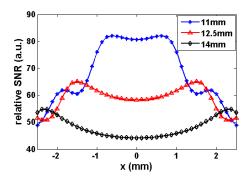


Figure 6. The relative SNR along x axis on the optimal plane of three RF coils of Figure 5 calculated from the simulation.

Figure 7. The orientations of  $B_0$  (blue) and  $B_1$  (red) on the optimal plane.

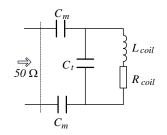


Figure 8. The tuning and matching circuit of the RF coil.

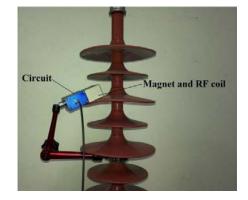


Figure 9. A photo of the UMR sensor (including the nonmagnetic clamp) measuring the shed of the composite insulator sample.

The impedance of the optimal RF coil was 0.486 + j11.18 ohms. A circuit (shown in Figure 8) was connected to the RF coil, and the composition of them was the RF probe. In order to guarantee the best excitation and reception, the resonance frequency of the RF probe was tuned to 5.017 MHz, and the impedance of the RF probe was matched to 50 ohms. For our RF coil, the capacitances  $C_m$  and  $C_t$  were 310 pF and 1875 pF, respectively.

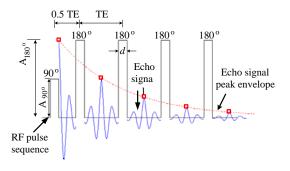
The optimized UMR sensor measuring the shed of the composite insulator sample is shown in Figure 9. The weight of the whole sensor (including the nonmagnetic clamp) is 0.556 kg and, the size is  $13 \text{ cm} \times 14 \text{ cm} \times 5 \text{ cm}$  which is compact and portable.

### 3. EXPERIMENTAL RESULTS AND DISCUSSIONS

All measurements were carried out with a Kea2 console (Magritek, Wellington, New Zealand), connected to a RF power amplifier (TOMCO Technologies, Australia) under the room temperature.

The test samples were the sheds of the composite insulators (used for 2 years, 8 years and 11 years) under 110 kv lines from the same area. Before the measurements, the surfaces of the samples were cleaned with detergent and dried. The employed pulse sequence was CPMG sequence [23] shown in Figure 10, where d was the pulse width, TE was the echo time,  $A_{180^{\circ}}$  and  $A_{90^{\circ}}$  were the power amplitude of 180° pulse and 90° pulse.

In all the measurements, the parameters of CPMG sequence were kept the same as follows: d is 6 µs, TE was 140 µs, the number of echoes was 3000, the repetition time was 1.5 s,  $A_{90^{\circ}}$  was -20 dB and  $A_{180}^{\circ}$  was -14 dB. For UMR experiment, due to the inhomogeneity of  $B_0$ , the peak envelop of CPMG echoes (Figure 10) has exponential decrease with the efficient transverse relaxation time  $T_{2eff}$ .



50 40 40 30 20 10 0 32 64 128 256 512 1024 Number of scans

Figure 10. CPMG sequence employed for the measurements.

Figure 11. The SNR vs the number of scans.

Since the strength of  $B_0$  of UMR is usually low, and the quantity of the excited hydrogen is small, the number of scans of the measurements requires to be more than traditional NMR to get acceptable SNR. Theoretically the larger the number of scans is, the larger the SNR will be. But increasing the number of scans will also increase the measurement time, and long measurement time is undesirable in practical engineering. In order to research the relationship between the number of scans of the measurements and the SNR, we employed composite insulator shed serviced for 11 years as the sample. The CPMG measurements were taken with different numbers of scans, and the ratio of the average value of the first ten echo-peaks to the average value of the last ten echo-peaks was defined as SNR. Figure 11 shows the relationship between SNR and number of scans. It can be observed in Figure 11 that the increment of SNR was not much with the number of scans more than 128. So the number of scans was set to be 128 to get good SNR as well as acceptable measurement time.

The composite insulator samples were detected for three times respectively. Figure 12 shows the peaks of CPMG echoes from the composite insulator used for 11 years.

The measured echo amplitudes  $M_n$  of the CPMG sequence can be expressed in terms of the distribution of relaxation times [24],  $f(T_{2eff})$ , by

$$M_n = \int dT_{2eff} f(T_{2eff}) e^{-nTE/T_{2eff}} + Noise$$
(2)

where TE was the echo time and n the number of echo. The extraction of the distribution  $f(T_{2eff})$  from the experiment data  $M_n$  is an ill-conditioned problem. The data processing of all the measurements was performed with the Contin program [25] which uses an inverse Laplace transform to yield the  $T_{2eff}$ distribution. The regularization parameter  $\alpha$  was determined to be 1.32 from the *L*-curve method. Figure 13 shows the  $T_{2eff}$  distributions of the three insulator samples. It can be observed that the  $T_{2eff}$ distribution curve of each sample has one peak which means only one main  $T_{2eff}$  of each sample. We considered the peak of the  $T_{2eff}$  distribution curve was the  $T_{2eff}$  of each sample.

According to the results from the inverse Laplace transform, for the sake of simplicity, the CPMG decays of all the measurements were analyzed in terms of a single exponential fit (red curve in Figure 12):

$$M(t) = Ae^{-t/T_{2eff}} + Noise \tag{3}$$

The results of all the measurements are demonstrated in Table 1. Figure 14 shows the normalized

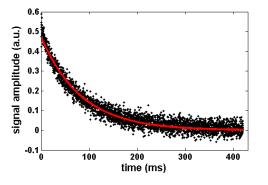


Figure 12. The peaks (black dot) of the CPMG echoes obtained from the shed of the composite insulator used for 11 years, and the red curve is single exponential decay fitted from the peaks.

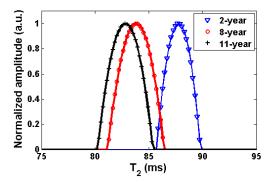


Figure 13. The  $T_2$  distribution obtained by the inverse Laplace transform of the three samples.

Table 1. The results of all the measurements.

Service time	Averaged $T_{2eff}$	Variation of $T_{2eff}$
2 years	$90.1\mu s$	$\pm 0.98\mu s$
8 years	$83.8\mu s$	$\pm 1.49\mu s$
11 years	$82.5\mu s$	$\pm 0.67\mu { m s}$

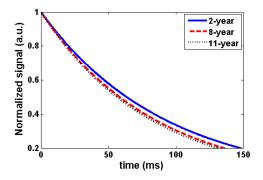


Figure 14. The exponential decay curves of the three insulator samples with normalized signal amplitude.

Table 2.  $T_{2eff}$  of three simples obtained by single exponential fitting and inverse Laplace transform.

Service time	$T_{2eff}$ -single exponential fitting	$T_{2eff}$ -inverse Laplace transform
2 years	$90.1\mu s$	87.7 μs
8 years	$83.8\mu s$	$83.8\mu \mathrm{s}$
11 years	$82.5\mu \mathrm{s}$	$82.8\mu s$

amplitude of the fitted exponential decay curves of the three samples. It can be obtained intuitively that the  $T_{2eff}$  of the samples decreases with the service time.

In order to compare the  $T_{2eff}$  obtained from single exponential fitting with the  $T_{2eff}$  obtained from inverse Laplace transform more clearly, we recorded them in Table 2. It can be observed that the results obtained from the single exponential fit and the inverse Laplace transform agree well with each other, and the  $T_{2eff}$  from both methods decreases with the service time of the sample. Especially the result of 8-year and 11-year samples were almost the same.

## 4. CONCLUSIONS

The mini UMR sensor designed in this paper is small, light weight and can achieve nondestructive testing at project site to judge the aging status of composite insulator sheds. The effective transverse relation time  $T_{2eff}$  tested by the sensor decreased with the service time of the insulators. The results encourages deep research of more samples from different environments.

Although the sensor designed in this paper can measure the transverse relation time quickly, the static magnetic field may be affected by the temperature. It is a barrier to obtain a stable effective transverse relation time of the sample at the project site where the temperature varies with time. The future work will focus on improving the temperature stability of the static magnetic field of the sensor by purchasing low temperature coefficient magnets or adding negative temperature coefficient magnetic materials on the magnets.

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