A *p*-Variable Higher-Order Finite Volume Time Domain Method for Electromagnetic Scattering

Avijit Chatterjee^{*} and Subodh Joshi

Abstract—A higher-order accurate solution to electromagnetic scattering problems is obtained at reduced computational cost in a *p*-variable finite volume time domain method in a scattered field formulation. Spatial operators of lower order, including first-order accuracy, are employed locally in substantial parts of the computational domain during the solution process. The use of computationally cheaper and lower order spatial operators does not affect the overall higher-order accuracy of the solution. The order of the spatial operator at a candidate cell during numerical simulation can vary in space and time and is dynamically chosen based on an order of magnitude comparison of scattered and incident fields at the cell centre. Numerical results are presented for electromagnetic scattering from perfectly conducting two-dimensional scatterers subject to transverse magnetic and transverse electric illumination.

1. INTRODUCTION

Numerical simulation of electromagnetic (EM) scattering problems with full wave solvers in time domain like Finite Difference Time Domain (FDTD) or Finite Volume Time Domain (FVTD) method can result in long simulation times. This is especially true while dealing with large electrical sizes $(a/\lambda \gg 1)$ where a is the characteristic size or with scatterers containing cavities and intakes. A major contributor to large simulation time is the requirement of small mesh sizes to adequately resolve spatial phenomenon. Mesh sizes are usually dictated by points-per-wavelength (PPW) criteria which may be heuristic in nature. Higher-order spatially accurate schemes are able to resolve spatial variations with lower PPW in the computational domain than lower order representations. Higher-order spatially accurate methods can achieve similar accuracy levels on much coarser discretization than lower-order methods. However, higher-order spatially accurate methods tend to be more expensive on a per-grid-point basis than lower order counterparts which mitigates some of the advantages accruing from the use of coarser meshes. Thus, there is significant motivation in developing computationally low cost higher-order numerical methods for solving the time-domain Maxwell's equations.

Multigrid (MG) methods [1, 2] based on cycling the numerical solution through a hierarchy of approximations either in space (h) or in polynomial order (p) or a combination of both have been used commonly to accelerate convergence to steady state of boundary value problems. *h*-MG methods are common in both finite volume and finite element frameworks while *p*-MG methods tend to be mostly restricted to mostly finite element framework [3]. Local *h* or *p* refinements have long been used, including for solving initial value problems, if the length scales to be resolved are not uniform across the computational domain and can cut down significantly on total computational time [4, 5]. Local refinement in the polynomial order (p) is again mostly restricted to finite element discretizations. A finite volume based solution of linear hyperbolic PDEs, including time-domain Maxwell's equations,

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^{*} Corresponding author: Avijit Chatterjee (avijit@aero.iitb.ac.in).

The authors are with the Department of Aerospace Engineering, Indian Institute of Technology Bombay, Mumbai, India.

by cycling through successive lower order p-approximations while retaining highest-order accuracy was proposed in [6, 7].

In the current work we propose a *p*-variable finite volume framework with an emphasis on solving EM scattering problems in the time domain. In the proposed framework, the time domain Maxwell equations which form a set of coupled linear hyperbolic PDEs, involving scattered field variables, are solved on a fixed grid but with the spatial operator formally varying in accuracy over the computational domain. The harmonic steady state solution obtained retains desired higher-order accuracy in spite of significant and not fixed parts of computational domain, processed using spatial operators of lower including first-order accuracy, during the simulation. The choice of accuracy of the spatial operator, done dynamically, is based on an order of magnitude comparison between the scattered and incident field at the cell center which serves as the measure of the local vertical scaling of the scattered field variable compared to the incident field. The framework requires an unified access to spatial operators of various orders of accuracy. For the present work the Essentially Non-Oscillatory (ENO) [8,9] methodology is used to locally obtain spatial operators of the desired accuracy, but it would be possible to use any higher-order numerical method for solving linear hyperbolic PDEs that similarly provides unified access to spatial operators of varying accuracy. Numerical results are presented for EM scattering from perfectly conducting circular cylinder and airfoil.

2. P-VARIABLE HIGHER-ORDER ACCURACY

Consider the scalar advection equation to be the scalar representation of the time-domain Maxwell's equations in differential form in a scattering process. The scalar advection equation for the scattered field is written as

$$\frac{\partial u}{\partial t} + c \frac{\partial u}{\partial x} = 0 \tag{1}$$

with wave speed $c \geq 0$. We assume u to represent a scattered field variable with

$$U = U_i + u \tag{2}$$

where U and U_i respectively represent the corresponding total and incident fields. All variables in Equation (1) can be nondimensionalized as $u^* = u/U_i$, $x^* = x/\lambda$ and $t^* = t/T$. λ and T are the wavelength and time period for the harmonic incident wave. Further all nondimensional values u^* , x^* and t^* lie in [0, 1] and in terms of order of magnitude are assumed to be O(1). Equation (1) can be written in corresponding nondimensional form as

$$\frac{\partial u^*}{\partial t^*} + \frac{\partial u^*}{\partial x^*} = 0. \tag{3}$$

The proposed *p*-variable method utilizes spatial operators of formally different orders of accuracy $(p \leq m)$ depending on the order of magnitude by which the local scattered field u(x,t) has been vertically scaled but always retains a local truncation error corresponding to the the highest *m*th order accuracy. Spatial operators of *m* and (m - 1)th order formal accuracy result in local truncation errors of similar magnitude when being applied respectively to scattered field function with vertical scaling that differ by one-order-of-magnitude. This fact can be used recursively to involve even lower order operators while retaining formal *m*th order accuracy. We show this using the nondimensional form and an order-of-magnitude analysis of the local truncation error.

Consider the scattered field function $u^*(x,t)$, discretization of the space derivative in Equation (3) with an m^{th} order accurate spatial operator results in a truncation error with leading term given by [10]

$$a(\Delta x^*)^m \frac{\partial^{m+1} u^*}{\partial^{m+1} x^*} \tag{4}$$

where a is a rational number. In a practical finite difference type formulation approximately 10 PPW or more would be required for a reasonable resolution for EM scattering problems which makes Δx^* at least one order of magnitude less than the representative wavelength λ . Thus, $\Delta x^* \sim O(1/10)$ in terms of order of magnitude. We now consider a scattered field function $v^*(x,t)$ obtained by vertically scaling $u^*(x,t)$ such that

$$v^*(x,t) \sim \Delta x^* \times u^*(x,t). \tag{5}$$

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Discretizing the space derivative in the nondimensionalized advection equation for $v^*(x,t)$ given by

$$\frac{\partial v^*}{\partial t^*} + \frac{\partial v^*}{\partial x^*} = 0, \tag{6}$$

with an (m-1)th order accurate spatial operator will similarly lead to a truncation error with leading term

$$b(\Delta x^*)^{m-1} \frac{\partial^m v^*}{\partial^m x^*} \sim b(\Delta x^*)^{m-1} \frac{\partial^m \Delta x^* u^*}{\partial^m x^*}.$$
(7)

For vertical scaling by constant Δx^* ,

$$\frac{\partial^m \triangle x^* u^*}{\partial^m x^*} \sim \triangle x^* \frac{\partial^m u^*}{\partial^m x^*} \tag{8}$$

is valid, using which Equation (7) can be rewritten as

$$b(\Delta x^*)^{m-1} \frac{\partial^m \Delta x^* u^*}{\partial^m x^*} \sim b(\Delta x^*)^m \frac{\partial^m u^*}{\partial^m x^*}.$$
(9)

We assume

$$\frac{\partial^m u^*}{\partial^m x^*} = \frac{\partial^{m+1} u^*}{\partial^{m+1} x^*} = O(1) \tag{10}$$

since u^* and x^* are both O(1) in the nondimensionalization process. This is similar to fluid mechanics boundary layer theory, where the nondimensional velocity and distance in the streamwise direction are both O(1), resulting in the first and second derivatives of the streamwise velocity in the streamwise direction also being O(1) [11]. Thus,

$$b(\Delta x^*)^{m-1} \frac{\partial^m \Delta x^* u^*}{\partial^m x^*} \sim b(\Delta x^*)^m \frac{\partial^{m+1} u^*}{\partial^{m+1} x^*}.$$
(11)

This implies the leading term of the truncation error resulting from spatial operators of the *m*th and (m-1)th accuracy given respectively in Equations (4) and (7) to be of comparable magnitude, i.e.,

$$a(\Delta x^*)^m \frac{\partial^{m+1} u^*}{\partial^{m+1} x^*} \sim b(\Delta x^*)^{m-1} \frac{\partial^m v^*}{\partial^m x^*}.$$
(12)

This can be applied recursively to bring in spatial operators of even lower order of accuracy while locally yielding spatial accuracy comparable to the highest *m*th order accuracy. Based on this a *p*variable algorithm can be constructed to obtain inexpensively a spatially higher-order accurate steady state solution for a scattering process. The algorithm for the *m*th order accuracy in a cell centered Finite Volume Time Domain (FVTD) framework can be of the form described below and can be easily included in an existing higher-order solver for EM scattering problems,

- If the cell centered scattered variable $u(x_j, t_k) \ge \Delta x \times U_i(x_j, t_k)$ the spatial operator is of order m. x_j, t_k respectively represent discrete space and time.
- For cell centered variable $\triangle x^{n+2} \times U_i(x_j, t_k) \le u(x_j, t_k) < \triangle x^{n+1} \times U_i(x_j, t_k)$ the spatial operator is of order m (n+1) with $n \ge 0$

with $U_i(x,t)$ assumed to be of similar order of magnitude throughout the domain and $\Delta x^n \times U_i(x_j,t_k) \sim U_i(x_j,t_k)/10^n$.

The above algorithm is used to cheaply obtain higher-order accurate solutions to EM scattering problems in an FVTD framework. A method of lines approach decouples the time and space discretizations, and the spatial discretization is obtained using an Essentially Non-Oscillatory (ENO) method which allows easy access to varying orders of spatial accuracy. The current implementation is in the ENO-Roe form [8, 9], which efficiently implements the ENO reconstruction based on the numerical fluxes instead of the cell averaged state variables and is described for the scalar law. Equation (1) is written as a scalar hyperbolic conservation law

$$u_t + f(u)_x = 0, (13)$$

has the spatial derivative at the *i*th grid point approximated as

$$\frac{\partial f(u)}{\partial x}|_{i} = \frac{1}{\Delta x} (\overline{f}_{i+1/2} - \overline{f}_{i-1/2}) + \mathsf{O}(\Delta x^{p})$$
(14)

where Δx is the grid size, p the order of the scheme, and $\overline{f}_{i+1/2}$ the numerical flux function at the right cell-face. The rth order accurate reconstruction of the numerical flux in the ENO scheme is

$$\overline{f}_{i+1/2} = \sum_{l=0}^{r-1} \alpha_{k,l}^r f_{i-r+1+k+l}$$
(15)

where $\alpha_{k,l}^r$ are the reconstruction coefficients, and k is the stencil index selected among the r candidate stencils. The stencil S_k can be written as

$$S_k = (x_{i+k-r+1}, x_{i+k-r+2}, \dots, x_{i+k})$$
(16)

and is locally the smoothest possible stencil. Details regarding reconstruction coefficients and stencil selection for ENO schemes are easily available in literature including [8,9]. Extension to the multidimensional system of equations like the time-domain Maxwell's equations can be obtained by decoupling the system into three scalar hyperbolic conservation laws normal to the cell faces [6, 12, 13].

The time domain Maxwell's equations for transverse magnetic (TM) or transverse electric (TE) waves are written in a conservative form [12]

$$\frac{\partial \boldsymbol{u}}{\partial t} + \frac{\partial \boldsymbol{f}(\boldsymbol{u})}{\partial x} + \frac{\partial \boldsymbol{g}(\boldsymbol{u})}{\partial y} = \boldsymbol{s}.$$
(17)

The vectors in Equation (17) for the TM waves are

$$\boldsymbol{u} = \begin{pmatrix} B_x \\ B_y \\ D_z \end{pmatrix}, \ \boldsymbol{f} = \begin{pmatrix} 0 \\ -D_z/\varepsilon \\ -B_y/\mu \end{pmatrix}, \ \boldsymbol{g} = \begin{pmatrix} D_z/\varepsilon \\ 0 \\ B_x/\mu \end{pmatrix} \boldsymbol{s} = \begin{pmatrix} 0 \\ 0 \\ -J_{iz} \end{pmatrix}$$
(18)

while that for the TE waves are

$$\boldsymbol{u} = \begin{pmatrix} B_z \\ D_x \\ D_y \end{pmatrix}, \ \boldsymbol{f} = \begin{pmatrix} D_y/\varepsilon \\ 0 \\ B_z/\mu \end{pmatrix}, \ \boldsymbol{g} = \begin{pmatrix} -D_x/\varepsilon \\ -B_z/\mu \\ 0 \end{pmatrix} \boldsymbol{s} = \begin{pmatrix} 0 \\ -J_{ix} \\ -J_{iy} \end{pmatrix}.$$
(19)

where **B** is the magnetic induction, **E** the electric field vector, **D** the electric field displacement, **H** the magnetic field vector, and $\mathbf{J}_{\mathbf{i}}$ the impressed current density vector, $\mathbf{D} = \varepsilon \mathbf{E}$, $\mathbf{B} = \mu \mathbf{H}$ with ε and μ respectively the permittivity and permeability in free space. The FVTD method solves the conservative Maxwell's equation in the integral form. Here a scattered field formulation is employed with the incident field assumed to be a solution of the Maxwell's equations in free space and scattered field variables solved for. Integrating the differential form of the conservation law, represented by Equation (17), in the absence of a source term over an arbitrary control volume Ω

$$\frac{\partial \int_{\Omega} \boldsymbol{u} d\boldsymbol{\mathcal{V}}}{\partial t} + \int_{\Omega} \boldsymbol{\nabla} .(\boldsymbol{F}(\boldsymbol{u})) d\boldsymbol{\mathcal{V}} = 0.$$
⁽²⁰⁾

F is the flux vector with components **f**, **g** in the Cartesian x, y directions. The integral form of the conservation law to be discretized is obtained by applying the divergence theorem as

$$\frac{\partial \int_{\Omega} \boldsymbol{u} d\boldsymbol{\mathcal{V}}}{\partial t} + \oint_{\mathcal{S}} \boldsymbol{F}(\boldsymbol{u}) \cdot \hat{\boldsymbol{n}} d\mathcal{S} = 0$$
(21)

with \hat{n} the outward unit normal vector. The spatially discretized form solved for in a scattered and cell-centered formulation in the present work is finally written as [12]

$$A_k \frac{d\boldsymbol{u}_k}{dt} + \sum_{j=1}^4 [(\boldsymbol{\mathcal{F}}(\boldsymbol{u}).\hat{\boldsymbol{n}}\boldsymbol{\mathcal{S}})_j]_k = 0$$
(22)

where the numerical flux $[(\mathcal{F}(u).\hat{n}S)_j]_k$ approximates the average flux through face j of cell k, and A_k represents the area of the quadrilateral cells in structured discretized space. In the present work, the semi-discretized form in Equation (22) is solved using higher-order ENO [8, 9] based spatial discretization described above and a second-order Runge-Kutta time integration. The ENO scheme is cast in a p-variable higher-order framework which results in highest (mth) order accurate solutions in the steady state, even while using spatial approximations with p < m based on an order of magnitude comparison of one or more selected field variables. The stability of the p-variable method is dictated by that of the underlying ENO method used to solve the linear hyperbolic PDEs and is given by the well-known Courant-Friedrichs-Lewy (CFL) condition.

3. NUMERICAL RESULTS

Numerical results are presented for EM scattering from 2D perfectly conducting circular cylinders as shown in Figure 1 and compared with the exact solution. A body confirming "O" mesh defines the computational domain with PEC boundary conditions on the cylinder surface and characteristic based far-field conditions at the outer boundary. Results are presented for both TM and TE continuous harmonic incident fields. Computations are performed for a fixed set of time periods of the incident harmonic wave, after which complex surface currents are obtained using a Fourier transform. The bistatic Radar Cross Section (RCS) or scattering width is then computed using a far-field transformation [14]. A discussion on the number of incident wave periods to be time-stepped for attaining sinusoidal steady state in a FDTD framework under harmonic incident excitation as attempted here is presented in [15]. The first problem considered is that of the circular cylinder subject to continuous harmonic incident TM illumination with $a/\lambda = 4.8$ where a is the cylinder radius and λ the wavelength of the incident wave [6, 12, 16]. Results are shown in terms of bistatic RCS and the absolute value of surface current after time stepping fixed incident time periods usually adequate for desired steady state response in such problems. Figures 2(a) and 2(b) show sample results for a conventional implementation for different spatial orders of accuracy on an "O" grid with 300 points in the circumferential direction corresponding to a resolution of 10 PPW on the scatterer surface after 5 time periods. The number of points in the radial direction is always kept constant at 50. A relatively lower resolution of 10 PPW on the cylinder surface is deliberately chosen to bring out the effect of the numerical discretization error on the solution obtained using different spatial orders of accuracy from the fourth to first. As expected, the highest fourth-order accurate solutions are closest to the exact solution with the first and second-order accurate solutions showing significant deviation away from near-specular-regions. The monostatic point is located at $\pm 180^{\circ}$ in the bistatic plot with 0° the perfect shadow. The same problem is now solved with a p-variable method with m = 4. An order of magnitude comparison of scattered and incident cellwise value of D_z is used to fix the local (cellwise) order of accuracy $(p \leq 4)$ of the spatial operator.



Figure 1. Schematic of a circular cylinder illuminated with an incident field.

Results are presented after 5 time periods in Figures 2(c) and 2(d) and compared with exact and conventional fourth-order results. Results from *p*-variable method with m = 4 match closely with conventional fourth-order results. L_1 , L_2 and L_{∞} norms of relative error for surface current density and bistatic RCS are tabulated in Table 1. It can be seen that the *p*-variable method shows similar relative errors to the conventional fourth-order ENO scheme.

The next problem considered is that of illumination by a continuous harmonic incident TE wave and $a/\lambda = 9.6$ [6, 12, 16]. The "O" grid with 600 points in circumferential direction is taken so that the resolution on the scatterer surface again corresponds to 10 PPW. Again, a deliberately coarse discretization is chosen to bring out the effect of spatial order of accuracy on the obtained solution. Figure 3(a) compares the bistatic RCS with the first, second, third and fourth-order accuracy after 5 time periods. The TE solution also starts deviating from the exact solution as formal spatial order of accuracy goes down, and this is especially apparent away from the near-specular-region. The problem is solved with a *p*-variable method and m = 4. The choice of spatial order *p* is based on an order of magnitude comparison of the scattered and incident values of B_z . Figure 3(b) compares the solution obtained with conventional fourth-order results. Again an almost exact match is obtained. Figure 4 shows the percentage of the computational domain over the entire simulation time processed by the first, second, third and fourth-order spatial operators while retaining an overall fourth-order accuracy for both TM and TE cases.



Figure 2. Bistatic RCS and surface current density $a/\lambda = 4.8$, continuous harmonic TM illumination, *p*-variable and conventional. (a) Surface current density. (b) Bistatic RCS. (c) Surface current density. (d) Bistatic RCS.

Table 1. Relative error norms for different numerical schemes including the *p*-variable method for bistatic RCS and surface current density in TM case.

Bistatic RCS										
Schemes	O(1)	O(2) ENO	O(3) ENO	O(4) ENO	p-variable					
L_1 Error	1.54E-01	7.05 E-02	2.12E-02	1.26E-02	1.28E-02					
L_2 Error	2.52E-01	1.24E-01	3.36E-02	1.70E-02	1.73E-02					
L_{∞} Error	3.89E-01	2.19E-01	6.28E-02	2.84 E-02	2.94E-02					
Surface Current Density										
L_1 Error	2.77E-01	1.09E-01	3.27E-02	2.02E-02	2.23E-02					
L_2 Error	2.50E-01	1.03E-01	2.64E-02	1.65 E-02	1.79E-02					
L_{∞} Error	2.41E-01	1.05E-01	2.61E-02	1.40E-02	1.54E-02					



Figure 3. Bistatic RCS $a/\lambda = 9.6$, continuous harmonic TE illumination. (a) Conventional (b) *p*-variable.



Figure 4. Computational work distribution for *p*-variable method. (a) TM case. (b) TE case.

We also consider scattering from a perfectly conducting NACA 0012 airfoil as shown in Figure 5(a). The airfoil chord length is 10 times the wavelength of the incident harmonic TM wave at broadside incidence [6, 12, 16]. Results are obtained using a body-fitted "O" grid with 200 points around the airfoil and 50 in the normal direction. Figure 6 compares RCS results after 5 time periods using regular fourth-order spatial accuracy and *p*-variable fourth-order (m = 4). Both results are compared with a "reference solution" obtained using regular fourth-order spatial accuracy but on a much finer grid with 1600 points around the airfoil and time stepped for 10 time periods. Again, like in the case of the circular cylinder an almost exact match is obtained between the conventional and *p*-variable method of the same formal accuracy. Figure 6 lists the percentage of the computational domain over time processed by spatial operators $p \leq 4$. The trend is similar to that for scattering from perfectly conducting circular cylinders.

Let the computing cost per-cell associated with the pth order scheme be C_p . For a p-variable



Figure 5. Bistatic RCS $a/\lambda = 10$, continuous harmonic TM illumination, *p*-variable and conventional. (a) NACA 0012 airfoil schematic. (b) Bistatic RCS.



Figure 6. Computational work distribution for *p*-variable method; airfoil case.

Table 2. Saving in computing time with *p*-variable method (m = 4).

Computational Performance — Work Units										
	TM Case			TE Case						
Cycles	Conventional	p-variable	%	Conventional	p-variable	%				
	O(4)	Method	Saving	O(4)	Method	Saving				
2	4.52e08	2.84e08	37.24	3.72e08	1.91e08	48.62				
5	9.04e08	6.36e08	29.62	7.47e08	4.40e08	41.11				
10	1.66e09	1.24e09	25.14	1.37e09	8.78e08	35.96				

method with m = 4, total computing cost (C_{total}) can then be written as,

$$C_{total} = \sum_{p=1}^{4} C_p n_p \tag{23}$$

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where, C_p is the computational cost per-cell at *p*th level and n_p the total number of cells processed at *p*th level. On the other hand, the uniformly 4th-order accurate scheme will incur a cost of $(n_T C_4)$ work units, where n_T is the total number of cells on the domain. The variation in computing cost with order of accuracy for a 2D FV ENO scheme is seen to follow $C_3 = 2.2C_2$ and $C_4 = 3.4C_2$ [17]. Assuming $C_2 = 2C_1$, we get $C_2/C_1 = 2, C_3/C_1 = 4.4$ and $C_4/C_1 = 6.8$. Table 2 shows the saving in computational cost over conventional fourth-order method in terms of work units assuming $C_1 = 1$ unit for test cases involving circular cylinders. A similar calculation yields a saving of 43.4% for the test case of NACA 0012 airfoil.

4. CONCLUSION

Desired higher-order spatial accuracy can be maintained, while using lower-order spatial operators in substantial parts of the computational domain in a *p*-variable FVTD method for solving EM scattering problems. Lower-order spatial operators come at much reduced computational cost and can cut down considerably on simulation time while retaining desired higher-order accuracy using the present method. An order of magnitude comparison of scattered and incident cell-centered EM field variables is used to decide on the local order of accuracy of the spatial operator. The local spatial order of accuracy can vary in space and time, and the proposed method can be easily integrated with existing higher-order FVTD techniques. The current implementation uses the ENO family to access spatial operators of desired order of accuracy as dictated by the order of magnitude comparison. Results are presented for the canonical case of EM scattering from a perfectly conducting circular cylinder as well as that of an airfoil.

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