Mutual Coupling Reduction in Microstrip Array Antenna by Employing Cut Side Patches and EBG Structures

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Abstract—This paper presents the simultaneous application of Minkowski fractal geometry and EBG structures for mutual coupling reduction in microstrip array antennas for the first time. In this approach, a modified version of Minkowski fractal geometry is applied on the patch elements, and at the same time 1D electromagnetic bandgap (EBG) structures, composed of 4 EBG elements, are placed between the array elements in a very close distance. Unlike many other coupling reduction methods, which have at least one of the issues of gain reduction or complex fabrication, the proposed method does need any via or double-sided etching and slightly increases the gain of the antenna, while an excellent reduction level of 22.7 dB has been achieved. To verify the concept, 2 array antennas with the spacing of λ_0 and $\lambda_0/3$ were fabricated and tested, showing very good agreement between predicted and measured results.

1. INTRODUCTION

High-gain antennas are highly desirable in long-distance communications such as satellite-terrestrial links and radar systems because of their highly-directive radiation patterns and their high signal to noise ratio. Perhaps the most traditional high-gain antennas are reflector antennas and horn antennas; however, their large non-planar configurations are not desired in space limited applications. Array antennas do not necessarily have such volumetric constraints, but they can become laterally oversized, due to large distance between array elements to avoid mutual coupling. Microstrip array antennas are one of the most attractive arrays because of their high efficiency, low weight, easy manufacture, and low cost [1]. They generally comprises several low-gain patch antennas arranged to form an array antenna. In many applications, such as wide-angle scanning required in satellite communication systems, the electrical spacing between the array elements needs to be minimal, resulting in unwanted mutual coupling between array elements, leading to degraded radiation efficiency [2]. Moreover, mutual coupling deteriorates the active input impedance of each array and causes scan blindness [3].

To overcome this problem in microstrip array antennas, various methods, such as Split-Ring Resonators (SRR) [4], concave patches [5], waveguide-based metasurfaces [6], and defected ground structures (DGS) [7] have been presented. Although these techniques may provide noticeable mutual coupling reduction, they may have some drawbacks. The replacement of conventional patches with concave patches can slightly improve the mutual coupling through a straightforward approach. However, this approach slightly reduces the gain of each element, due to the reduction of effective aperture of each element [5].

In order to achieve a greater reduction (about 10 dB) in mutual coupling, without affecting the gain of the antenna, SRR can be used [4]. The limitation associated to this method is the complex

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prototyping for higher frequencies, when the resonators dimensions are extremely sensitive, and any tolerance in fabrication can affect the resonance frequency. Another approach to achieve a good level mutual coupling isolation at a price of more complex design is the use of DGS. As reported in [7], mutual coupling and gain of the antenna are reduced by around 20 dB and 1.4 dB, respectively by introducing four DGS cells. Unlike most of the above-mentioned techniques which have a by-product of slight gain-reduction, metasurfaces can be utilized for mutual coupling reduction. Metasurfaces are 2D form of metamaterials and have exceptional abilities for controlling electromagnetic waves, which can be used for variety of applications, such as suppressing unwanted coupling effects, beamsteering [6, 8]. In [6], a waveguide based metasurface was designed to exhibit a bandgap with two transmission zeros, which attributed to negative permeability and negative permittivity in the vicinity of magnetic resonance and electric resonance, respectively. The mutual coupling of the antenna was reduced by around 11 dB, while the gain was increased by 0.5 dB, because of the metasurface introduction. In a different approach, artificial intelligence has recently been used in conjunction with electromagnetic simulators to solve some of EM problems including mutual coupling through an evolutionary approach [9–11].

The aim of this study is to propose a new size-effective approach for mutual reduction without sacrificing the gain of the antenna, which does not need double substrate etching or via holes. In the proposed approach for the first time, 1D EBG structures are used in conjunction with a geometrical modification, inspired by Minkowski fractal, on the patch elements. The proposed hybrid technique significantly reduces the number of EBG elements required to remarkably attenuate the mutual coupling between the patch elements by 22.7 dB, with 0.8 dB gain enhancement, reaching up to 5.2 dB. The envelope correlation coefficient of the proposed antenna system is less than 0.03 for 5.2 to 6 GHz, and the antenna efficiency is greater than 40% over the same frequency range. The size of the antenna is $80 \text{ mm} \times 150 \text{ mm}$. In addition, unlike [12] our proposed EBG consists of only four cells arranged in one column, resulting in smaller distance between the patches. This low profile antenna, like other low profile planar antennas, has a potential to be fabricated with flexible materials such as antenna based on polydimethylsiloxane and conductive fabric due to its high gain, high efficiency, and low profile [13, 14].

2. ANTENNA CONFIGURATION

2.1. EBG Unit Cell

The use of EBG structures for mutual coupling reduction has been widely investigated, and various types of EBG structures have been proposed for this application [15]. The concept of EBG arises from the optical domain, where photonic crystals can forbid the propagation of light or allow it in certain directions. This feature can be used in suppressing surface-waves and realization of EBG resonator antennas [15–18]. These periodic structures are composed of a metallic or dielectric element, mostly in the form of 2D and 3D structure geometries. The proposed structure in this study is a 1D EBG structure, where 4 elements are arranged along with one axis to suppress the mutual coupling between two patch elements. The proposed EBG structure is easily fabricable, without the need of any vias. The process of designing unit cell is explained as follows.

Step 1 (choosing unit cell structure): The EBG unit cell used for mutual coupling reduction is shown in Fig. 1. The idea of using EBG-based unit cells for mutual coupling reduction comes from [18], in which the unit cells are used for improving the bandwidth of Fabry-Perot antennas. The EBG structure can be modelled by an LC equivalent circuit if the periodicity of structure is small compared to the operating wavelength [19]. In order to have a compact unit cell with lower resonant frequency, a slot is added to the square unit cell.

Step 2 (investigating the unit cell band gap): After choosing the explained unit cell, its S-parameters are studied. The width and length of the slot is optimized through HFSS modelling simulator to tune the resonance at 5.6 GHz, where the highest return loss is achieved.

Step 3 (array of cells is placed between patches): After optimizing the dimensions of the unit cells operating at 5.6 GHz, the unit cells are arranged in an array of 4×1 and placed between two patches with high mutual coupling. Some small modification is required to cell dimension due to a frequency shift in this structure. The final dimensions of the unit cell are shown in Table 1.





Figure 1. The EBG unit cell.

Figure 2. The schematic of patches in proposed microstrip antenna.

Table 1.	Dimensions	of EBG	unit cell ((in mm)).
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w_t	l_t	w_s	l_s	t
16	14.2	10.4	6	1.2

2.2. Patch Configuration

In array structures, there is an electromagnetic interaction between patch elements, which is stronger across the adjacent sides. The effect of electromagnetic interaction between the elements can be reduced by applying a Minkowski-like fractal geometry on the patch as shown in Fig. 2.

Minkowski fractal was proposed by Hermann Minkowski in 1906 for the first time and brought into electromagnetic community in late 1990, and since then it has been used for modifying electromagnetic and microwave components [20]. In the proposed patch configuration, cuts are small compared to the patch size and would not affect the antenna performance. The shape and size of cuts are shown in Fig. 2. The introduction of the EBG structure along with the patch modification is depicted in Fig. 3.



Figure 3. Schematic of microstrip array antenna with the proposed EBG structure and new patches. (a) Top view, (b) side view.

3. SIMULATION RESULTS

The proposed antenna is implemented on an FR4 substrate with a thickness of 1.6 mm, relative permittivity of 4.8, and loss tangent of 0.015. The resonance frequency of the antenna is around 5.6 GHz suitable for WLAN applications. The proposed antenna is fed by two coaxial probes (see Fig. 3) and simulated using HFSS 14. The dimensions of the patches are listed in Table 2.

Table 2. Dimensions of the patches (in mm).

W_s	L_s	w_p	l_p	x	y	d
80	150	33.5	37.5	10	18	52

In order to give a better insight into the individual effects of the EBG structures as well as the modified patch elements, the mutual coupling of the system is analysed separately. As shown in Fig. 4, each individual technique has been effective in reducing the mutual coupling, showing 4 dB and 12 dB for the case of only EBG and only patch modifications, respectively, when the two patches are separated by λ_0 (52 mm). In the next step, both techniques are used at the same time to achieve greatly higher isolation level.

Figure 5 shows the mutual coupling for the conventional patches without the EBG structures and the proposed hybrid method (EBG and modified patches) for two different distances of λ_0 (52 mm) and $\lambda_0/3$ (19 mm) between the two patches. As can be seen from this figure, the mutual coupling between patches has been significantly reduced from 24 dB to about 44 dB for the case of $\lambda_0/3$ and from 26 dB to about 48 dB for the case of λ_0 distance between the two patches.

To further investigate the effects of EBG and cut patches, individually, the current distribution are shown in Fig. 6. As depicted in Fig. 6, there is strong current induced on the left-side patch, as a result of the unwanted coupling between the patches. This coupling is decreased by the introduction of the EBG structure, resulting in lower current density on the left-side patch, as shown in Fig. 6(b). This current density is then further reduced, when the fractal geometry is applied on the patches as depicted in Fig. 6(c).

Envelop correlation of the conventional and proposed antennas with a separation of $\lambda_0/3$ is shown in Fig. 7. In the antenna with modified patches and EBG cells, the ECC is less than 0.03 throughout the entire operating frequency band. In addition, as shown in Fig. 8, the antenna gain and efficiency are improved in the proposed structure compared to conventional antenna.

0dB



-10dB S11 w/o EBG, d2 S11 w cuts and EBG, d S11 w cuts and EBG, d2 -20dB S12 w/o EBG, dI S parameter (dB) S12 w/o EBG, d2 S12 w cuts and EBG -30dB 40dB -50dF -60dB 5.90 5.40 5.50 5.60 5.70 5.80 6.00 5.20 5.30 Freq (GHz)

Figure 4. Simulated S parameter of the antenna with cut side patches and EBG structure at the distance of λ_0 .

Figure 5. Simulated S parameter of conventional and proposed antennas at two distances of $\lambda_0/3(d1)$ and $\lambda_0(d2)$.



Figure 6. Surface current distributions at 5.6 GHz, (a) without EBG and cuts, (b) with EBG without cuts, (c) with EBG and cuts.



Figure 7. Simulated ECC of conventional and proposed antennas at distances of $\lambda_0/3(d1)$.



Figure 8. Simulated peak gain and radiation efficiency of conventional and proposed antennas at distances of $\lambda_0/3(d1)$.

4. MEASUREMENT RESULTS

In order to verify the simulation results, four prototypes of the conventional and proposed antennas at two distances of λ_0 and $\lambda_0/3$ are fabricated. Fig. 12 shows the fabricated prototypes, which is



Figure 9. Measurement results of conventional and proposed antennas at two distances of $\lambda_0/3(d1)$ and $\lambda_0(d2)$.



Figure 10. Simulated and measurement results of conventional and proposed antennas at distance of $\lambda_0/3$.



Figure 11. Simulated and measured far-field radiation patterns of conventional and proposed antennas in (a) *E*-plane (dB) and (b) *H*-plane.



Figure 12. Prototypes of cut side microstrip patches with EBG structure at two different distances $(\lambda_0, \lambda_0/3)$.

Table	3.	Comr	parison	of	the	proposed	antenna	characte	erization	with	previous	works
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Reference	Current work	[21]	[22]	[23]
Centre frequency (GHz)	$5.6\mathrm{GHz}$	$5.3\mathrm{GHz}$	$5.8\mathrm{GHz}$	$3.6\mathrm{GHz}$
Complexity of unit cell in fabrication	Simple	Simple	Simple	complex
Via	No	Yes	Yes	Yes
Number of unit cells	4	10	7	5
Gain improvement	$0.8\mathrm{dB}$	NA	No change	$1.2\mathrm{dB}$
Gain (dBi)	5.1	5	NA	11.5
Mutual coupling reduction	$22.7\mathrm{dB}$	20.5	$10\mathrm{dB}$	-18
Envelope Correlation Coefficient	0.03	NA	NA	NA
Efficiency	> 40%	NA	NA	NA
Size (λ_0^2)	2.8×1.5	1.2×0.6	1.3×0.81	0.93×1.27
Bandwidth	2%	3.9%	2.5%	5.5%

a combination of EBG structures and fractal geometry, for the separation distance of 19 mm (top figure) and 52 mm (bottom figure). The S-parameters are measured using Agilent-N5230A PNA-L series network analyser and are shown in Fig. 9. It is depicted that the measurement results are in good agreement with the predicted results from simulations (see Fig. 10). The radiation patterns in E-plane and H-plane of the array are presented in Fig. 11. As expected, radiation characterization after reduction of mutual coupling in the antenna is enhanced, and gain of the structure is improved by 0.8 dB.

For demonstrating the advantages of this proposed structure compared to other works, the properties of this structure and some previous works are listed in Table 3. As shown, the proposed antenna has better mutual coupling reduction, without imposing further complexity to the antenna system and without any via and extra layers.

5. CONCLUSION

This paper presents an effective mutual coupling reduction method for patch arrays, in which two different techniques are combined, for the first time, to achieve a significant reduction level of around 22.7 dB, through a cost and size-effective mechanism. In this method, Minkowski-like fractal geometry is applied on the patch elements, and simultaneously 1D EBG elements are placed very close to the patches. Unlike most of mutual reduction methods, the proposed approach not only does not reduce the gain, but also causes 0.8 dB increase in the gain of the antenna.

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